CURRENT-LIMITED, POWER-DISTRIBUTION SWITCHES

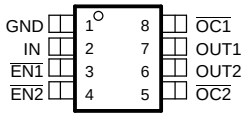
FEATURES

- **70-mΩ High-Side MOSFET**
- **1.5-A Continuous Current**
- **Thermal and Short-Circuit Protection**
- **Accurate Current Limit (1.6 A min, 2.6 A max)**
- **Operating Range: 2.7 V to 5.5 V**
- **0.6-ms Typical Rise Time**
- **Undervoltage Lockout**
- **Deglitched Fault Report ($\overline{OC}$)**
- **No $\overline{OC}$ Glitch During Power Up**
- **1-μA Maximum Standby Supply Current**
- **Bidirectional Switch**
- **Ambient Temperature Range: 0°C to 70°C**
- **ESD Protection**
- **UL Listed - File No. E169910**

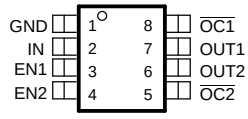
APPLICATIONS

- **Heavy Capacitive Loads**
- **Short-Circuit Protections**

TPS2060
DGN PACKAGE
(TOP VIEW)



TPS2064
DGN PACKAGE
(TOP VIEW)



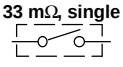
DESCRIPTION

The TPS206x power-distribution switches are intended for applications where heavy capacitive loads and short-circuits are likely to be encountered. This device incorporates 70-mΩ N-channel MOSFET power switches for power-distribution systems that require multiple power switches in a single package. Each switch is controlled by a logic enable input. Gate drive is provided by an internal charge pump designed to control the power-switch rise times and fall times to minimize current surges during switching. The charge pump requires no external components and allows operation from supplies as low as 2.7 V.

When the output load exceeds the current-limit threshold or a short is present, the device limits the output current to a safe level by switching into a constant-current mode, pulling the overcurrent ($\overline{OCx}$) logic output low. When continuous heavy overloads and short-circuits increase the power dissipation in the switch, causing the junction temperature to rise, a thermal protection circuit shuts off the switch to prevent damage. Recovery from a thermal shutdown is automatic once the device has cooled sufficiently. Internal circuitry ensures that the switch remains off until valid input voltage is present. This power-distribution switch is designed to set current limit at 2.1 A typically.

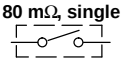
GENERAL SWITCH CATALOG

33 mΩ, single



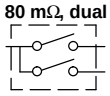
TPS201xA	0.2 A − 2 A
TPS202x	**0.2 A − 2 A**
TPS203x	0.2 A − 2 A

80 mΩ, single



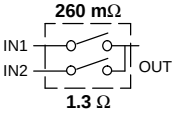
TPS2014	600 mA
TPS2015	1 A
TPS2041B	500 mA
TPS2051B	500 mA
TPS2045	250 mA
TPS2055	250 mA
TPS2061	1 A
TPS2065	1 A

80 mΩ, dual



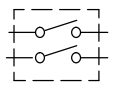
TPS2042B	500 mA
TPS2052B	500 mA
TPS2046	250 mA
TPS2056	250 mA
TPS2062	1 A
TPS2066	1 A
TPS2060	1.5 A
TPS2064	1.5 A

260 mΩ



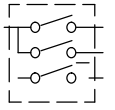
TPS2100/1	
IN1	500 mA
IN2	10 mA
TPS2102/3/4/5	
IN1	500 mA
IN2	100 mA

80 mΩ, dual



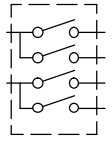
TPS2080	500 mA
TPS2081	500 mA
TPS2082	500 mA
TPS2090	250 mA
TPS2091	250 mA
TPS2092	250 mA

80 mΩ, triple



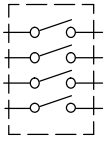
TPS2043	500 mA
TPS2053B	500 mA
TPS2047	250 mA
TPS2057	250 mA
TPS2063	1 A
TPS2067	1 A

80 mΩ, quad



TPS2044B	500 mA
TPS2054B	500 mA
TPS2048	250 mA
TPS2058	250 mA

80 mΩ, quad



TPS2085	500 mA
TPS2086	500 mA
TPS2087	500 mA
TPS2095	250 mA
TPS2096	250 mA
TPS2097	250 mA

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

AVAILABLE OPTION AND ORDERING INFORMATION

T_A	ENABLE	RECOMMENDED MAXIMUM CONTINUOUS LOAD CURRENT	TYPICAL SHORT-CIRCUIT CURRENT LIMIT AT 25°C	NUMBER OF SWITCHES	PACKAGED DEVICES (1)(2)
					MSOP (DGN)
0°C to 70°C	Active low	1.5 A	2.1 A	Dual	TPS2060DGN
	Active high				TPS2064DGN

(1) The package is available taped and reeled. Add an R suffix to device types (e.g., TPS2060DGN).
(2) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted(1)

			UNIT
V_I	Input voltage range, $V_{I(IN)}$(2)		-0.3 V to 6 V
	Input voltage range, $V_{I(/ENx)}$, $V_{I(ENx)}$		-0.3 V to 6 V
	Voltage range, $V_{I(/OC)}$, $V_{I(OCx)}$		-0.3 V to 6 V
V_O	Output voltage range, $V_{O(OUT)}$(2), $V_{O(OUTx)}$		-0.3 V to 6 V
I_O	Continuous output current, $I_{O(OUT)}$, $I_{O(OUTx)}$		Internally limited
	Continuous total power dissipation		See Dissipation Rating Table
T_J	Operating virtual junction temperature range		0°C to 105°C
T_{stg}	Storage temperature range		-65°C to 150°C
	Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds		260°C
	Electrostatic discharge (ESD) protection	Human body model MIL-STD-883C	2 kV
		Charge device model (CDM)	500 V

(1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages are with respect to GND.

DISSIPATING RATING TABLE

PACKAGE	T_A < 25°C POWER RATING	DERATING FACTOR ABOVE T_A = 25°C	T_A = 70°C POWER RATING
DGN-8	1712.3 mW	17.123 mW/°C	941.78 mW

RECOMMENDED OPERATING CONDITIONS

		MIN	MAX	UNIT
V_I	Input voltage, $V_{I(IN)}$	2.7	5.5	V
	Input voltage, $V_{I(ENx)}$, $V_{I(/ENx)}$	0	5.5	V
I_O	Continuous output current, $I_{O(OUTx)}$	0	1.5	A
T_J	Operating virtual junction temperature	0	105	°C

ELECTRICAL CHARACTERISTICS

over recommended operating junction temperature range, $V_{I(IN)}$ = 5.5 V, I_O = 1 A, $V_{I(/ENx)}$ = 0 V, or $V_{I(ENx)}$ = 5.5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS(1)			MIN	TYP	MAX	UNIT
POWER SWITCH								
$r_{DS(on)}$	Static drain-source on-state resistance, 5-V operation and 3.3-V operation	$V_{I(IN)}$ = 5 V or 3.3 V,	I_O = 1.5 A	0°C < T_J < 105°C		70	115	mΩ
	Static drain-source on-state resistance, 2.7-V operation(2)	$V_{I(IN)}$ = 2.7 V,	I_O = 1.5 A	0°C < T_J < 105°C		75	125	mΩ
t_r(2)	Rise time, output	$V_{I(IN)}$ = 5.5 V	C_L = 1 µF, R_L = 5 Ω	T_J = 25°C		0.6	1.5	ms
		$V_{I(IN)}$ = 2.7 V				0.4	1	
t_f(2)	Fall time, output	$V_{I(IN)}$ = 5.5 V			0.05		0.5	
		$V_{I(IN)}$ = 2.7 V			0.05		0.5	
ENABLE INPUT $\overline{EN}$ OR EN								
V_{IH}	High-level input voltage	2.7 V < $V_{I(IN)}$ < 5.5 V			2			V
V_{IL}	Low-level input voltage	2.7 V < $V_{I(IN)}$ < 5.5 V					0.8	
I_I	Input current	$V_{I(/ENx)}$ = 0 V or 5.5 V, $V_{I(ENx)}$ = 0 V or 5.5 V			-0.5		0.5	µA
t_{on}(3)	Turnon time	C_L = 100 µF, R_L = 5 Ω					3	ms
t_{off}(3)	Turnoff time	C_L = 100 µF, R_L = 5 Ω					10	
CURRENT LIMIT								
I_{OS}	Short-circuit output current	$V_{I(IN)}$ = 5 V, OUT connected to GND, device enabled into short-circuit			1.6	2.1	2.6	A
I_{OC_TRIP}(3)	Over-current trip threshold	$V_{I(IN)}$ = 5 V, Current ramp (≤ 100 A/s) on OUT				3.2	3.9	
Supply current, low-level output		No load on OUT, $V_{I(/ENx)}$ = 5.5 V, or $V_{I(ENx)}$ = 0 V		T_J = 25°C		0.5	1	µA
				0°C < T_J < 105°C		0.5	5	
Supply current, high-level output		No load on OUT, $V_{I(/ENx)}$ = 0 V, or $V_{I(ENx)}$ = 5.5 V		T_J = 25°C		50	70	µA
				0°C < T_J < 105°C		50	90	
Leakage current		OUT connected to ground, $V_{I(/ENx)}$ = 5.5 V, or $V_{I(ENx)}$ = 0 V		0°C < T_J < 105°C		1		µA
Reverse leakage current		$V_{I(OUTx)}$ = 5.5 V, IN = ground(3)		T_J = 25°C		0.2		µA
UNDERVOLTAGE LOCKOUT								
Low-level input voltage, IN					2		2.5	V
Hysteresis, IN		T_J = 25°C				75		mV
OVERCURRENT $\overline{OC1}$ and $\overline{OC2}$								
Output low voltage, $V_{OL(OCx)}$		$I_{O(/OCx)}$ = 5 mA					0.4	V
Off-state current(3)		$V_{O(/OCx)}$ = 5 V or 3.3 V					1	µA
$\overline{OC}$ deglitch(3)		$\overline{OCx}$ assertion or deassertion			4	8	15	ms
THERMAL SHUTDOWN(4)								
Thermal shutdown threshold(3)					135			°C
Recovery from thermal shutdown(3)					125			°C
Hysteresis(3)						10		°C

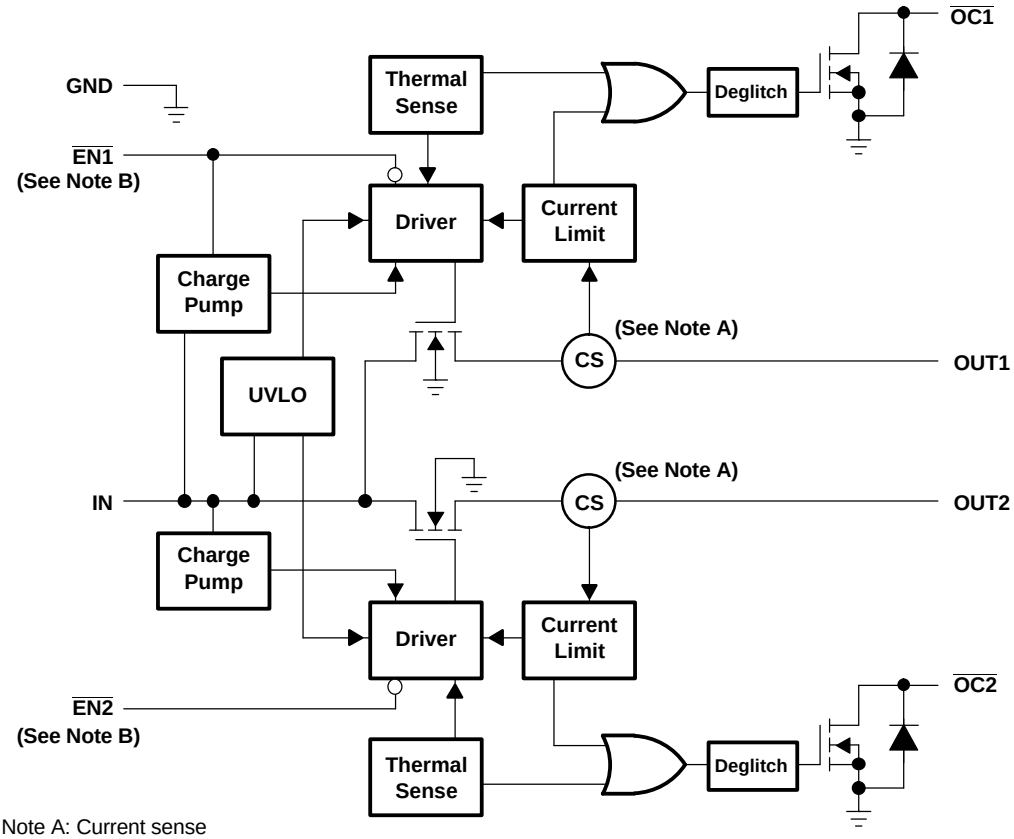
(1) Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.
(2) Not tested in production, specified by design.
(3) Not tested in production, specified by design.
(4) The thermal shutdown only reacts under overcurrent conditions.

DEVICE INFORMATION

Terminal Functions (TPS2060 and TPS2064)

TERMINAL			I/O	DESCRIPTION
NAME	NO.			
	TPS2060	TPS2064		
$\overline{EN1}$	3	-	I	Enable input, logic low turns on power switch IN-OUT1
$\overline{EN2}$	4	-	I	Enable input, logic low turns on power switch IN-OUT2
EN1	-	3	I	Enable input, logic high turns on power switch IN-OUT1
EN2	-	4	I	Enable input, logic high turns on power switch IN-OUT2
GND	1	1		Ground
IN	2	2	I	Input voltage
$\overline{OC1}$	8	8	O	Overcurrent, open-drain output, active low, IN-OUT1
$\overline{OC2}$	5	5	O	Overcurrent, open-drain output, active low, IN-OUT2
OUT1	7	7	O	Power-switch output, IN-OUT1
OUT2	6	6	O	Power-switch output, IN-OUT2

Functional Block Diagram



Note A: Current sense
Note B: Active low ($\overline{ENx}$) for TPS2060. Active high (ENx) for TPS2064.

PARAMETER MEASUREMENT INFORMATION

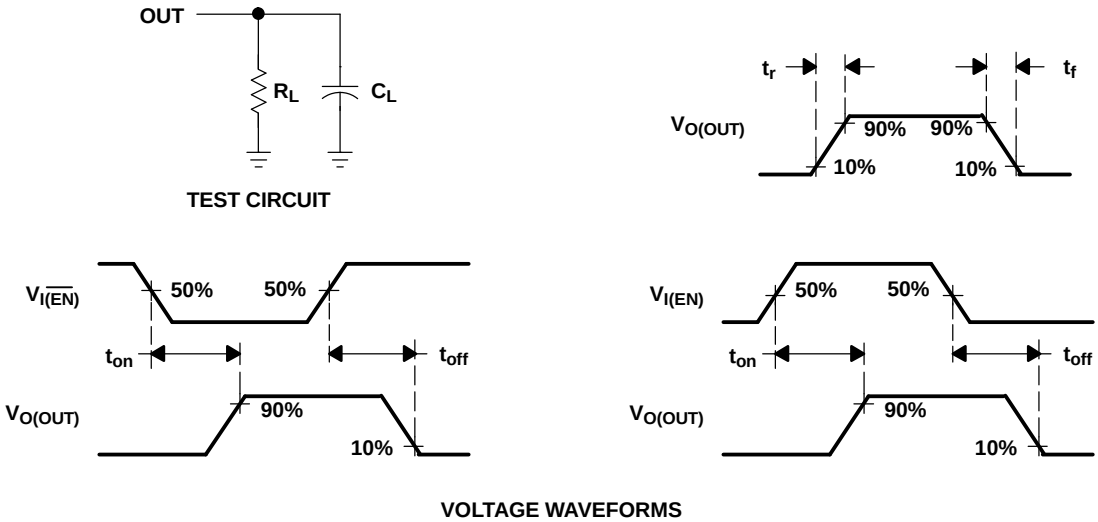


Figure 1. Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION (continued)

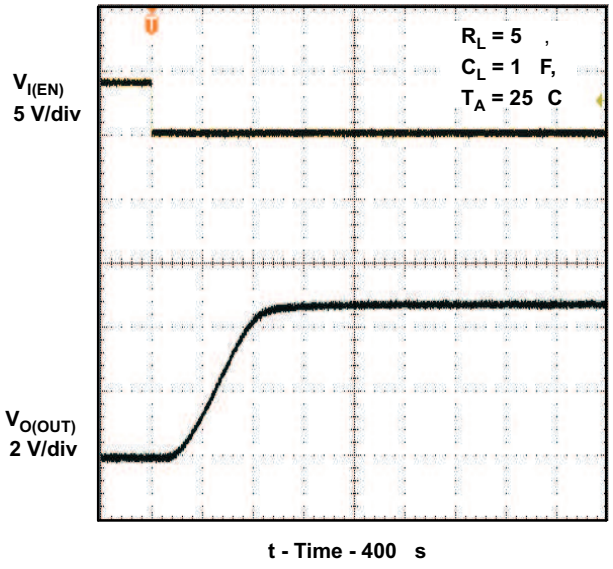


Figure 2. Turnon Delay and Rise Time With 1-μF Load

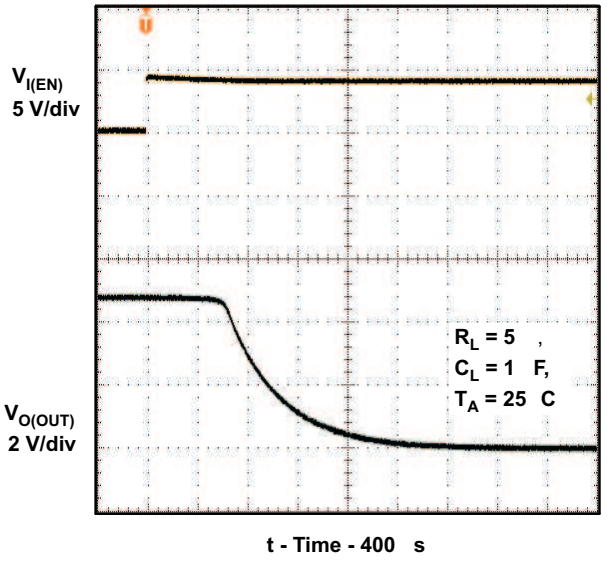


Figure 3. Turnoff Delay and Fall Time With 1-μF Load

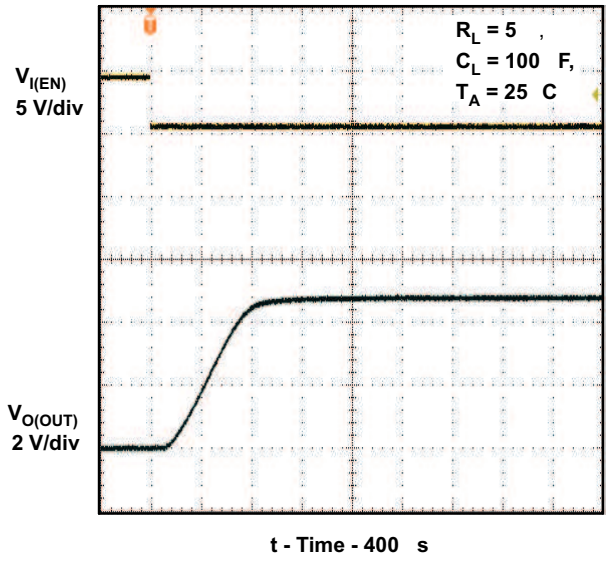


Figure 4. Turnon Delay and Rise Time With 100-μF Load

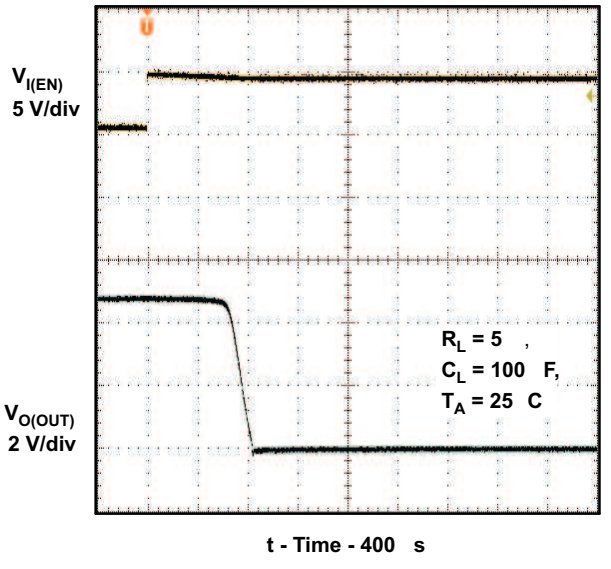


Figure 5. Turnoff Delay and Fall Time With 100-μF Load

PARAMETER MEASUREMENT INFORMATION (continued)

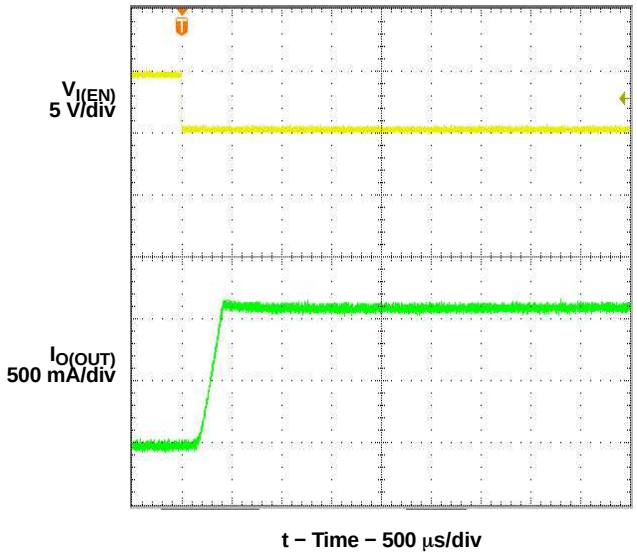


Figure 6. Short-Circuit Current, Device Enabled Into Short

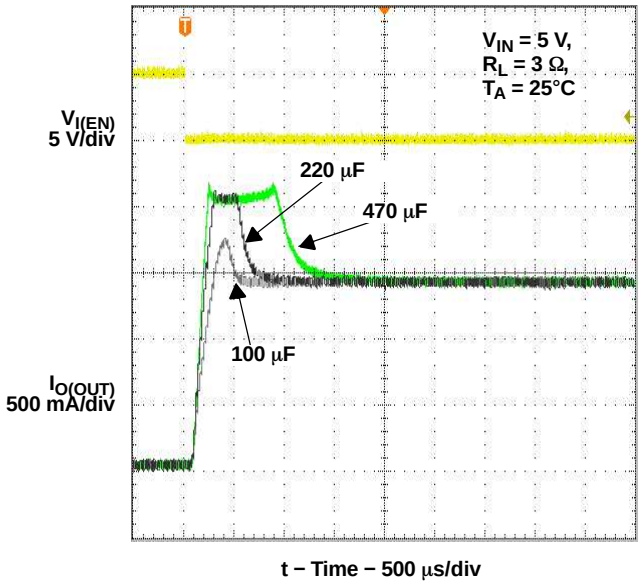


Figure 7. Inrush Current With Different Load Capacitance

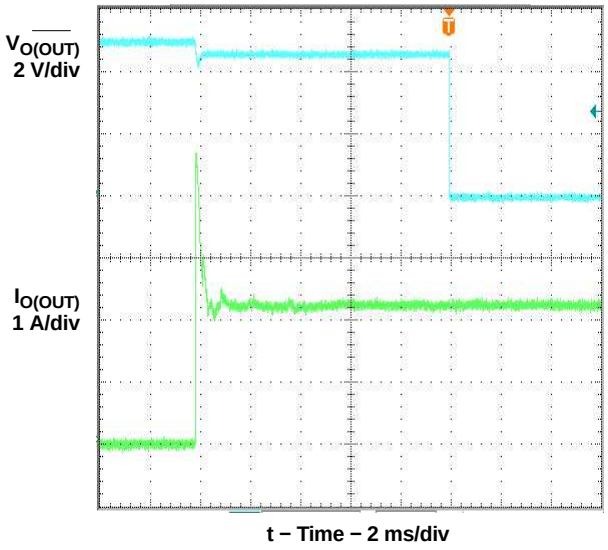


Figure 8. 0.6-Ω Load Connected to Enabled Device

TYPICAL CHARACTERISTICS

TURNON TIME vs INPUT VOLTAGE

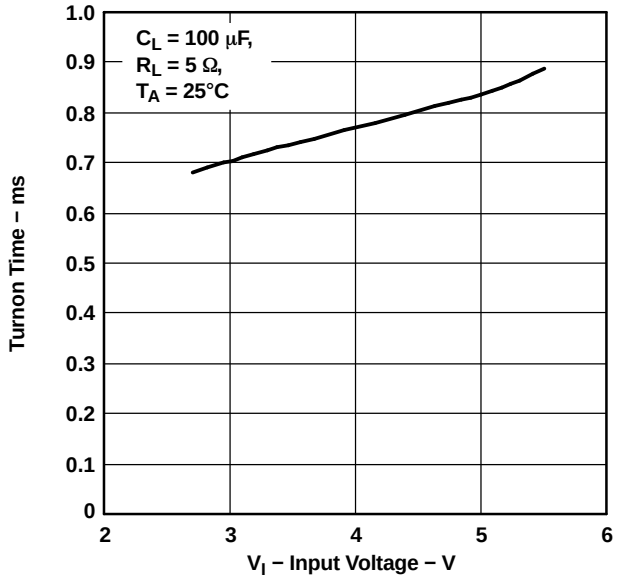


Figure 9.

TURNOFF TIME vs INPUT VOLTAGE

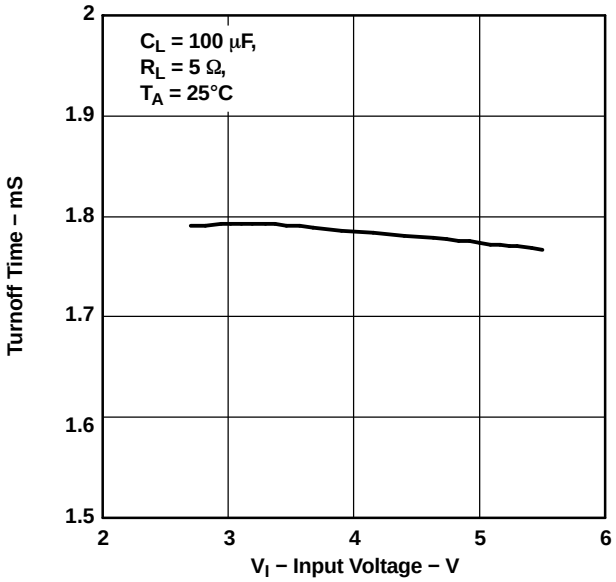


Figure 10.

RISE TIME vs INPUT VOLTAGE

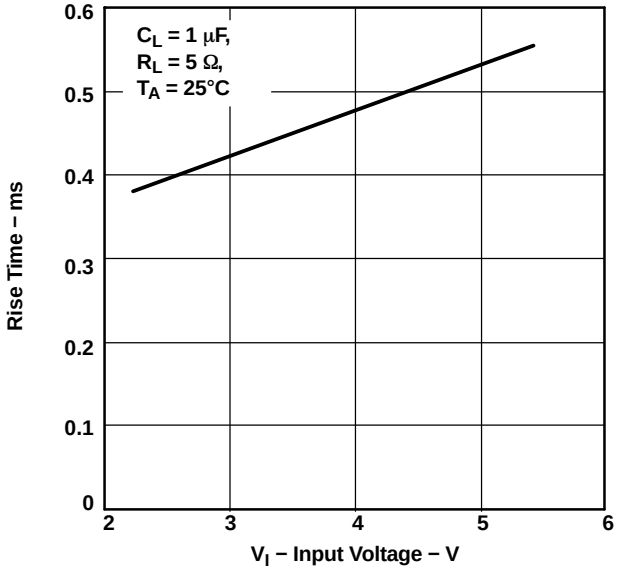


Figure 11.

FALL TIME vs INPUT VOLTAGE

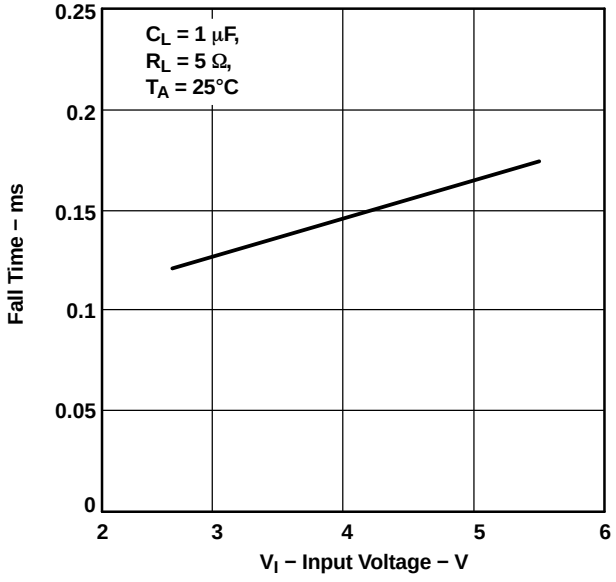


Figure 12.

TYPICAL CHARACTERISTICS (continued)

SUPPLY CURRENT, OUTPUT ENABLED
vs
JUNCTION TEMPERATURE

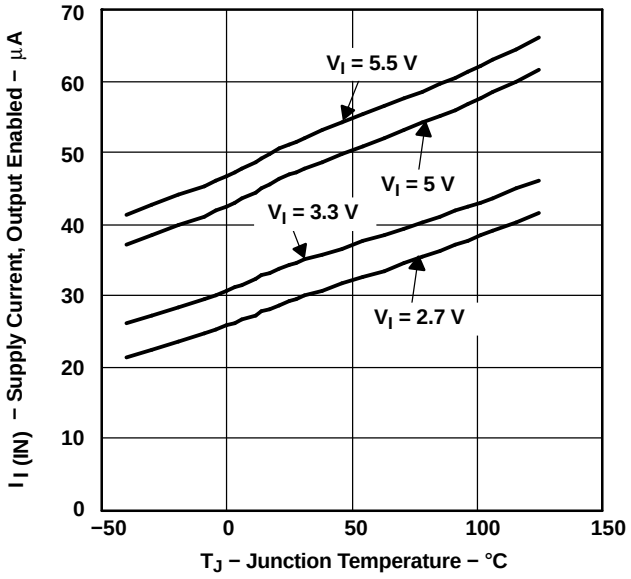


Figure 13.

SUPPLY CURRENT, OUTPUT DISABLED
vs
JUNCTION TEMPERATURE

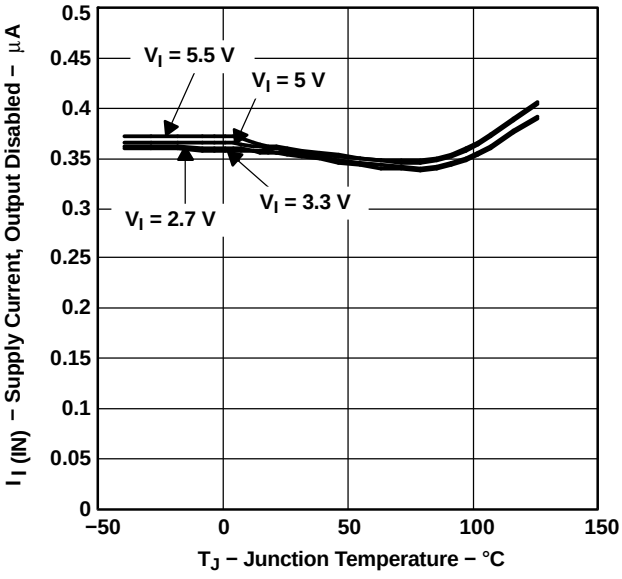


Figure 14.

STATIC DRAIN-SOURCE ON-STATE RESISTANCE
vs
JUNCTION TEMPERATURE

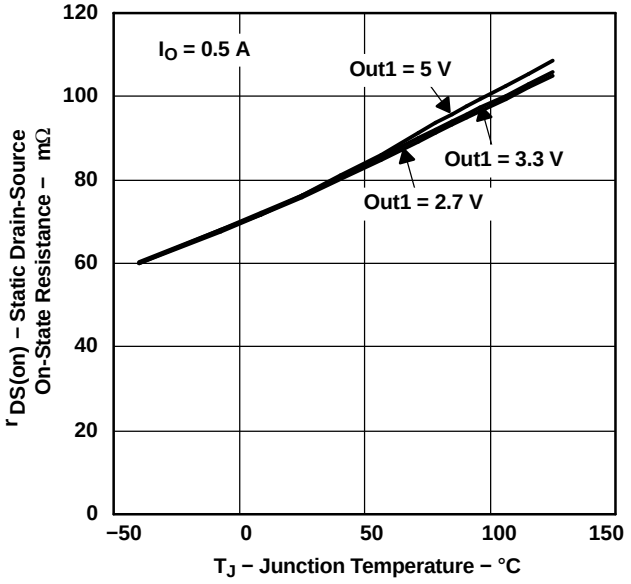


Figure 15.

SHORT-CIRCUIT OUTPUT CURRENT
vs
JUNCTION TEMPERATURE

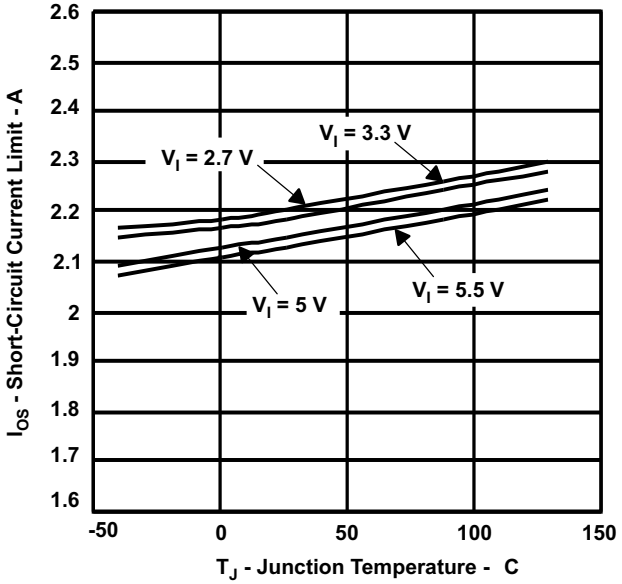


Figure 16.

TYPICAL CHARACTERISTICS (continued)

THRESHOLD TRIP CURRENT
vs
INPUT VOLTAGE

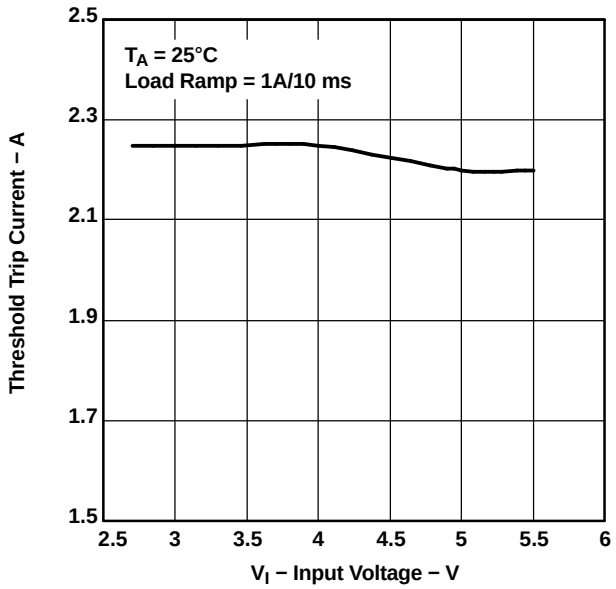


Figure 17.

UNDERVOLTAGE LOCKOUT
vs
JUNCTION TEMPERATURE

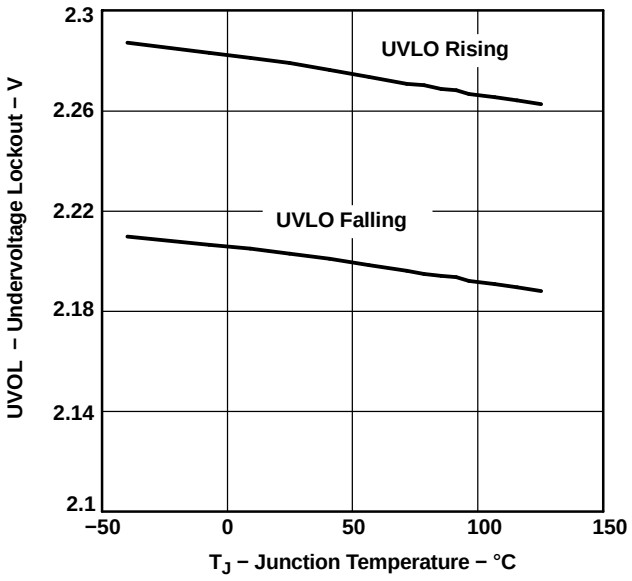


Figure 18.

CURRENT-LIMIT RESPONSE
vs
PEAK CURRENT

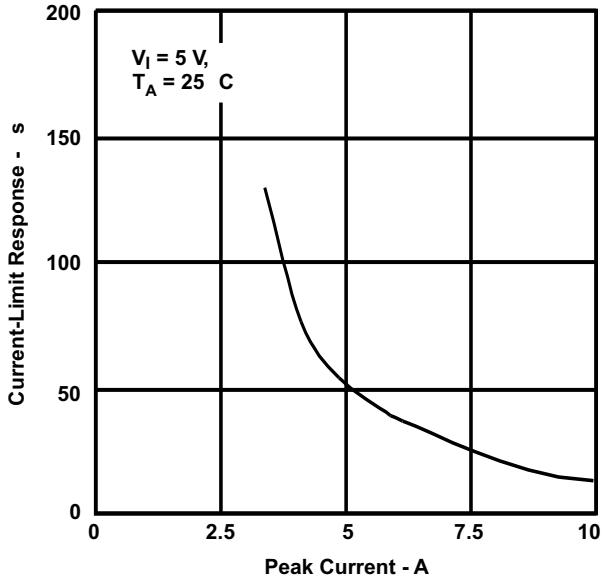


Figure 19.

APPLICATION INFORMATION

POWER-SUPPLY CONSIDERATIONS

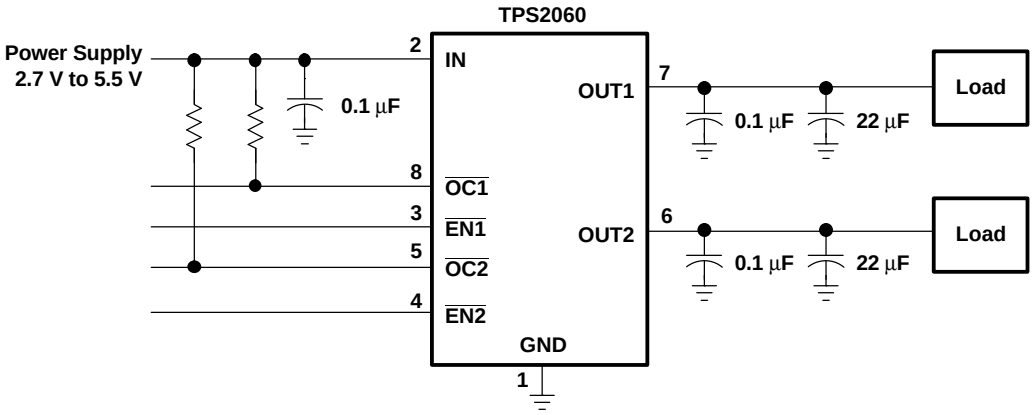


Figure 20. Typical Application

A 0.01-µF to 0.1-µF ceramic bypass capacitor between IN and GND, close to the device, is recommended. Placing a high-value electrolytic capacitor on the output pin(s) is recommended when the output load is heavy. This precaution reduces power-supply transients that may cause ringing on the input. Additionally, bypassing the output with a 0.01-µF to 0.1-µF ceramic capacitor improves the immunity of the device to short-circuit transients.

OVERCURRENT

A sense FET is employed to check for overcurrent conditions. Unlike current-sense resistors, sense FETs do not increase the series resistance of the current path. When an overcurrent condition is detected, the device maintains a constant output current and reduces the output voltage accordingly. Complete shutdown occurs only if the fault is present long enough to activate thermal limiting.

Three possible overload conditions can occur. In the first condition, the output has been shorted before the device is enabled or before $V_{I(IN)}$ has been applied (see Figure 13). The TPS206x senses the short and immediately switches into a constant-current output.

In the second condition, a short or an overload occurs while the device is enabled. At the instant the overload occurs, high currents may flow for a short period of time before the current-limit circuit can react. After the current-limit circuit has tripped (reached the overcurrent trip threshold), the device switches into constant-current mode.

In the third condition, the load has been gradually increased beyond the recommended operating current. The current is permitted to rise until the current-limit threshold is reached or until the thermal limit of the device is exceeded (see Figure 14). The TPS206x is capable of delivering current up to the current-limit threshold without damaging the device. Once the threshold has been reached, the device switches into its constant-current mode.

$\overline{\text{OC}}$ RESPONSE

The $\overline{\text{OCx}}$ open-drain output is asserted (active low) when an overcurrent or overtemperature shutdown condition is encountered after a 10-ms deglitch timeout. The output remains asserted until the overcurrent or overtemperature condition is removed. Connecting a heavy capacitive load to an enabled device can cause a momentary overcurrent condition; however, no false reporting on $\overline{\text{OCx}}$ occurs due to the 10-ms deglitch circuit. The TPS206x is designed to eliminate false overcurrent reporting. The internal overcurrent deglitch eliminates the need for external components to remove unwanted pulses. $\overline{\text{OCx}}$ is not deglitched when the switch is turned off due to an overtemperature shutdown.

APPLICATION INFORMATION (continued)

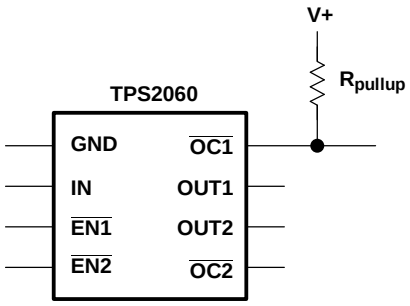


Figure 21. Typical Circuit for the $\overline{OC}$ Pin

POWER DISSIPATION AND JUNCTION TEMPERATURE

The low on-resistance on the N-channel MOSFET allows the small surface-mount packages to pass large currents. The thermal resistances of these packages are high compared to those of power packages; it is good design practice to check power dissipation and junction temperature. Begin by determining the $r_{DS(on)}$ of the N-channel MOSFET relative to the input voltage and operating temperature. As an initial estimate, use the highest operating ambient temperature of interest and read $r_{DS(on)}$ from Figure 15. Using this value, the power dissipation per switch can be calculated by:

- $P_D = r_{DS(on)} \times I^2$

Multiply this number by the number of switches being used. This step renders the total power dissipation from the N-channel MOSFETs.

Finally, calculate the junction temperature:

- $T_J = P_D \times R_{\theta JA} + T_A$

Where:

- T_A= Ambient temperature °C
- $R_{\theta JA}$ = Thermal resistance
- P_D = Total power dissipation based on number of switches being used.

Compare the calculated junction temperature with the initial estimate. If they do not agree within a few degrees, repeat the calculation, using the calculated value as the new estimate. Two or three iterations are generally sufficient to get a reasonable answer.

THERMAL PROTECTION

Thermal protection prevents damage to the IC when heavy-overload or short-circuit faults are present for extended periods of time. The TPS206x implements a thermal sensing to monitor the operating junction temperature of the power distribution switch. In an overcurrent or short-circuit condition, the junction temperature rises due to excessive power dissipation. Once the die temperature rises to approximately 140°C due to overcurrent conditions, the internal thermal sense circuitry turns the power switch off, thus preventing the power switch from damage. Hysteresis is built into the thermal sense circuit, and after the device has cooled approximately 10°C, the switch turns back on. The switch continues to cycle in this manner until the load fault or input power is removed. The $\overline{OCx}$ open-drain output is asserted (active low) when an overtemperature shutdown or overcurrent occurs.

UNDERVOLTAGE LOCKOUT (UVLO)

An undervoltage lockout ensures that the power switch is in the off state at power up. Whenever the input voltage falls below approximately 2 V, the power switch is quickly turned off. This facilitates the design of hot-insertion systems where it is not possible to turn off the power switch before input power is removed. The UVLO also keeps the switch from being turned on until the power supply has reached at least 2 V, even if the switch is enabled. On reinsertion, the power switch is turned on, with a controlled rise time to reduce EMI and voltage overshoots.

APPLICATION INFORMATION (continued)

UNIVERSAL SERIAL BUS (USB) APPLICATIONS

The universal serial bus (USB) interface is a 12-Mb/s, or 1.5-Mb/s, multiplexed serial bus designed for low-to-medium bandwidth PC peripherals (e.g., keyboards, printers, scanners, and mice). The four-wire USB interface is conceived for dynamic attach-detach (hot plug-unplug) of peripherals. Two lines are provided for differential data, and two lines are provided for 5-V power distribution.

USB data is a 3.3-V level signal, but power is distributed at 5 V to allow for voltage drops in cases where power is distributed through more than one hub across long cables. Each function must provide its own regulated 3.3 V from the 5-V input or its own internal power supply.

The USB specification defines the following five classes of devices, each differentiated by power-consumption requirements:

- Hosts/self-powered hubs (SPH)
- Bus-powered hubs (BPH)
- Low-power, bus-powered functions
- High-power, bus-powered functions
- Self-powered functions

SPHs and BPHs distribute data and power to downstream functions. The TPS206x has higher current capability than required by one USB port; so, it can be used on the host side and supplies power to multiple downstream ports or functions.

HOST/SELF-POWERED AND BUS-POWERED HUBS

Hosts and SPHs have a local power supply that powers the embedded functions and the downstream ports (see Figure 22). This power supply must provide from 5.25 V to 4.75 V to the board side of the downstream connection under full-load and no-load conditions. Hosts and SPHs are required to have current-limit protection and must report overcurrent conditions to the USB controller. Typical SPHs are desktop PCs, monitors, printers, and stand-alone hubs.

APPLICATION INFORMATION (continued)

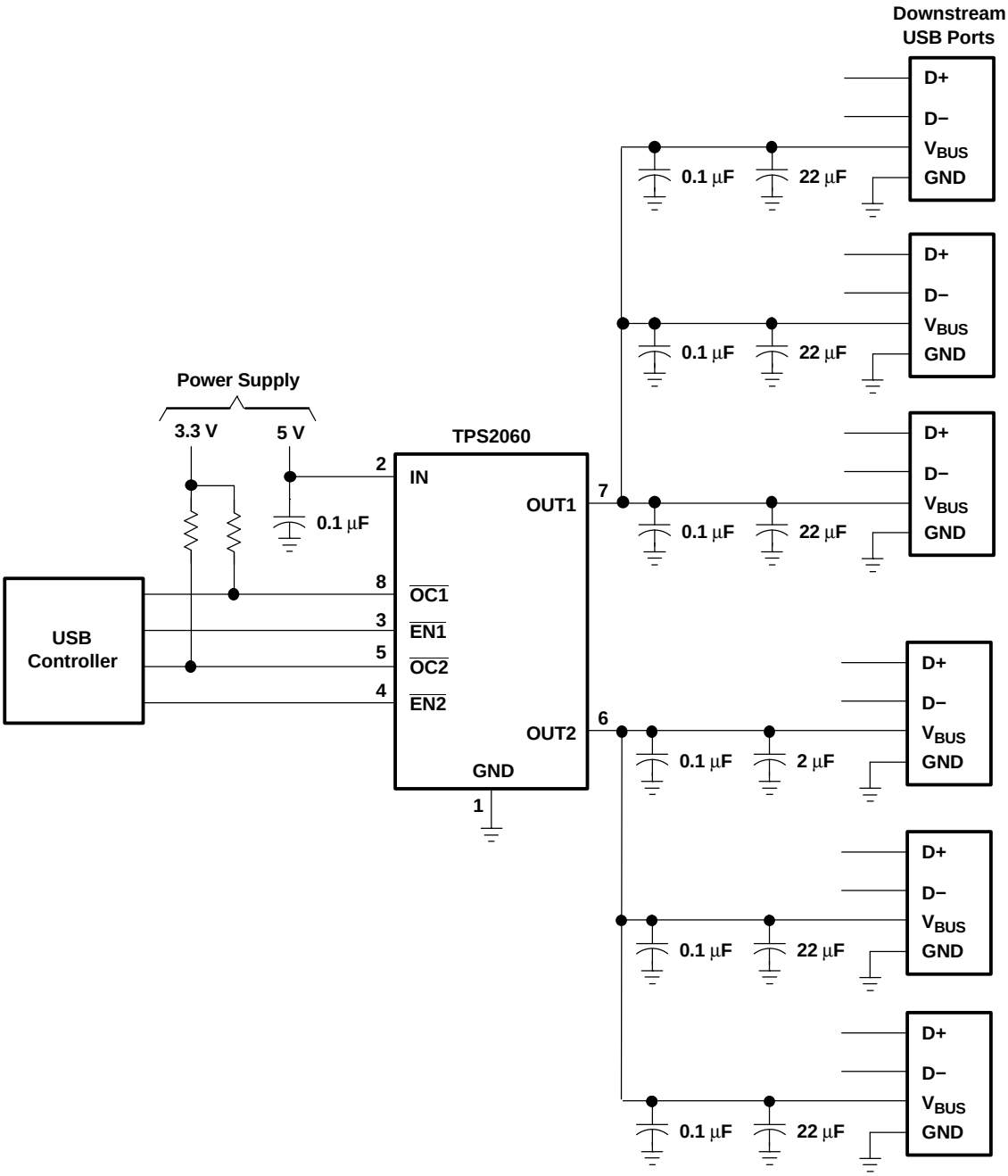


Figure 22. Typical Six-Port USB Host / Self-Powered Hub

BPHs obtain all power from upstream ports and often contain an embedded function. The hubs are required to power up with less than one unit load. The BPH usually has one embedded function, and power is always available to the controller of the hub. If the embedded function and hub require more than 100 mA on power up, the power to the embedded function may need to be kept off until enumeration is completed. This can be accomplished by removing power or by shutting off the clock to the embedded function. Power switching the embedded function is not necessary if the aggregate power draw for the function and controller is less than one unit load. The total current drawn by the bus-powered device is the sum of the current to the controller, the embedded function, and the downstream ports, and it is limited to 500 mA from an upstream port.

APPLICATION INFORMATION (continued)

LOW-POWER BUS-POWERED AND HIGH-POWER BUS-POWERED FUNCTIONS

Both low-power and high-power bus-powered functions obtain all power from upstream ports; low-power functions always draw less than 100 mA; high-power functions must draw less than 100 mA at power up and can draw up to 500 mA after enumeration. If the load of the function is more than the parallel combination of 44 Ω and 10 µF at power up, the device must implement inrush current limiting (see Figure 23). With TPS206x, the internal functions could draw more than 500 mA, which fits the needs of some applications such as motor driving circuits.

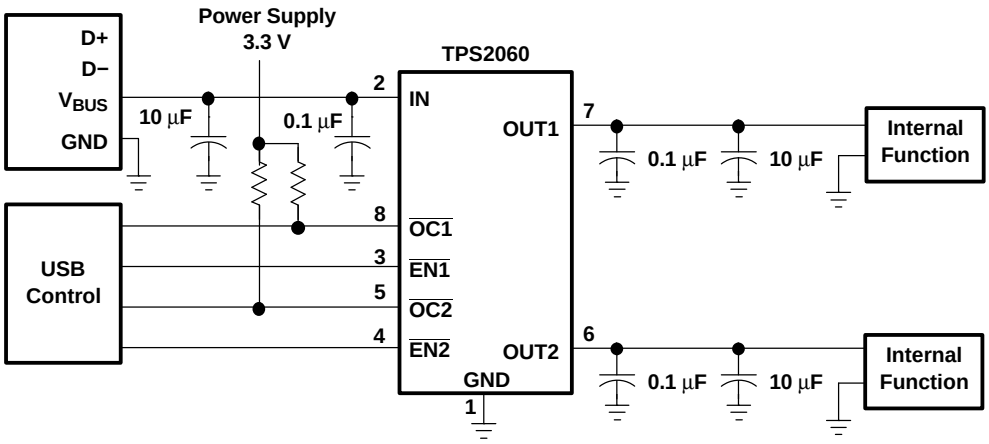


Figure 23. High-Power Bus-Powered Function

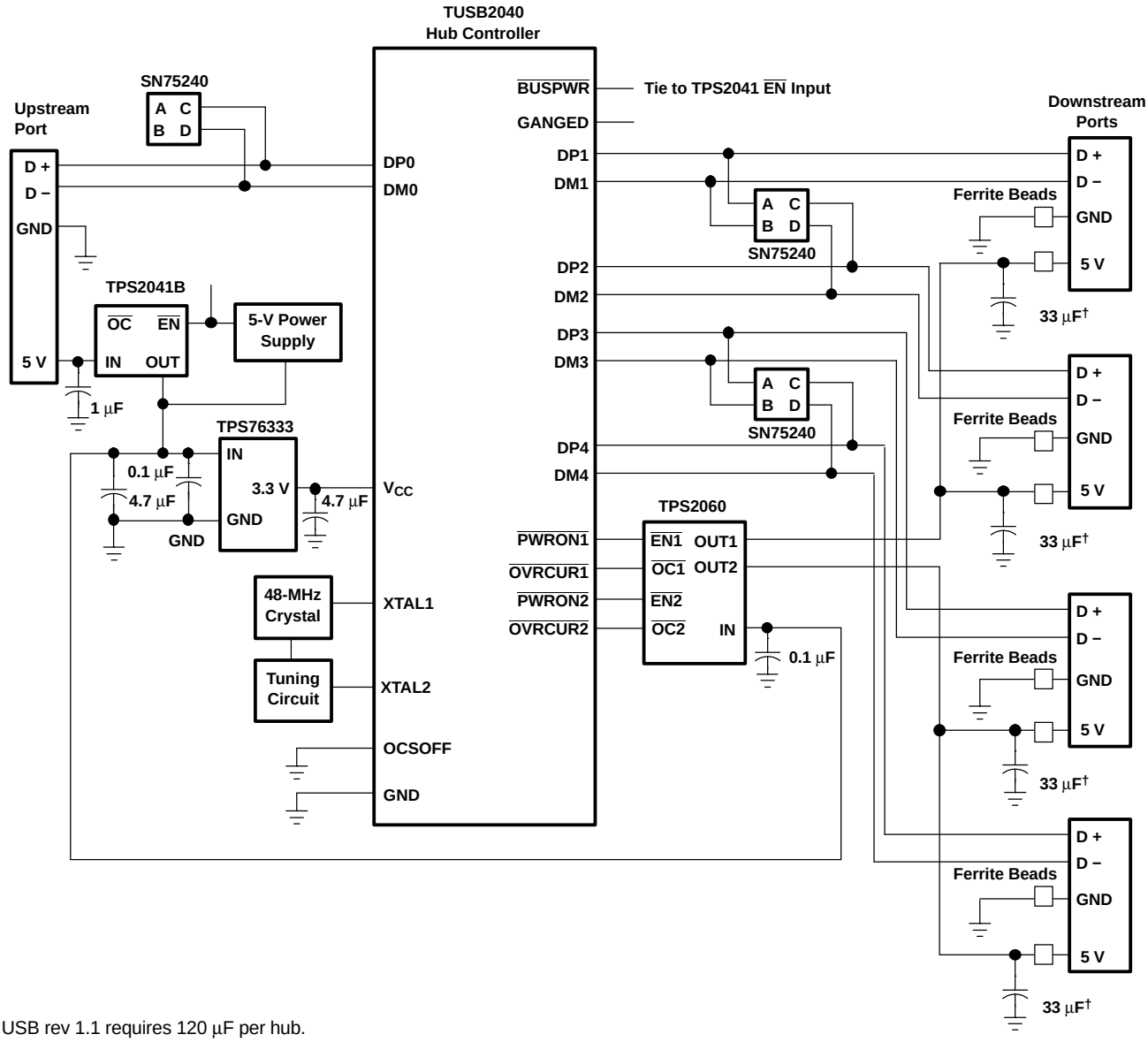
USB POWER-DISTRIBUTION REQUIREMENTS

USB can be implemented in several ways, and, regardless of the type of USB device being developed, several power-distribution features must be implemented.

- Hosts/SPHs must:
  - Current-limit downstream ports
  - Report overcurrent conditions on USB V_{BUS}
- BPHs must:
  - Enable/disable power to downstream ports
  - Power up at <100 mA
  - Limit inrush current (<44 Ω and 10 µF)
- Functions must:
  - Limit inrush currents
  - Power up at <100 mA

The feature set of the TPS206x allows them to meet each of these requirements. The integrated current-limiting and overcurrent reporting is required by hosts and self-powered hubs. The logic-level enable and controlled rise times meet the need of both input and output ports on bus-powered hubs, as well as the input ports for bus-powered functions (see Figure 24).

APPLICATION INFORMATION (continued)



† USB rev 1.1 requires 120 µF per hub.

Figure 24. Hybrid Self / Bus-Powered Hub Implementation

GENERIC HOT-PLUG APPLICATIONS

In many applications it may be necessary to remove modules or pc boards while the main unit is still operating. These are considered hot-plug applications. Such implementations require the control of current surges seen by the main power supply and the card being inserted. The most effective way to control these surges is to limit and slowly ramp the current and voltage being applied to the card, similar to the way in which a power supply normally turns on. Due to the controlled rise times and fall times of the TPS206x, these devices can be used to provide a softer start-up to devices being hot-plugged into a powered system. The UVLO feature of the TPS206x also ensures that the switch is off after the card has been removed, and that the switch is off during the next insertion. The UVLO feature insures a soft start with a controlled rise time for every insertion of the card or module.

APPLICATION INFORMATION (continued)

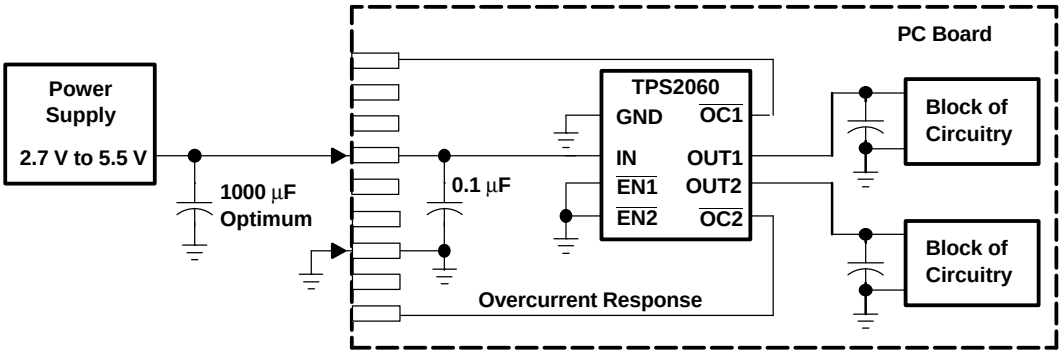


Figure 25. Typical Hot-Plug Implementation

By placing the TPS206x between the V_{CC} input and the rest of the circuitry, the input power reaches these devices first after insertion. The typical rise time of the switch is approximately 1 ms, providing a slow voltage ramp at the output of the device. This implementation controls system surge currents and provides a hot-plugging mechanism for any device.

DETAILED DESCRIPTION

Power Switch

The power switch is an N-channel MOSFET with a low on-state resistance. Configured as a high-side switch, the power switch prevents current flow from OUT to IN and IN to OUT when disabled. The power switch supplies a minimum current of 1.5 A.

Charge Pump

An internal charge pump supplies power to the driver circuit and provides the necessary voltage to pull the gate of the MOSFET above the source. The charge pump operates from input voltages as low as 2.7 V and requires little supply current.

Driver

The driver controls the gate voltage of the power switch. To limit large current surges and reduce the associated electromagnetic interference (EMI) produced, the driver incorporates circuitry that controls the rise times and fall times of the output voltage.

Enable ($\overline{ENx}$)

The logic enable disables the power switch and the bias for the charge pump, driver, and other circuitry to reduce the supply current. The supply current is reduced to less than 1 μA when a logic high is present on $\overline{ENx}$, or when a logic low is present on ENx. A logic zero input on $\overline{ENx}$, or a logic high input on ENx restores bias to the drive and control circuits and turns the switch on. The enable input is compatible with both TTL and CMOS logic levels.

Overcurrent ($\overline{OCx}$)

The $\overline{OCx}$ open-drain output is asserted (active low) when an overcurrent or overtemperature condition is encountered. The output remains asserted until the overcurrent or overtemperature condition is removed. A 10-ms deglitch circuit prevents the $\overline{OCx}$ signal from oscillation or false triggering. If an overtemperature shutdown occurs, the $\overline{OCx}$ is asserted instantaneously.

DETAILED DESCRIPTION (continued)

Current Sense

A sense FET monitors the current supplied to the load. The sense FET measures current more efficiently than conventional resistance methods. When an overload or short circuit is encountered, the current-sense circuitry sends a control signal to the driver. The driver in turn reduces the gate voltage and drives the power FET into its saturation region, which switches the output into a constant-current mode and holds the current constant while varying the voltage on the load.

Thermal Sense

The TPS206x implements a thermal sensing to monitor the operating temperature of the power distribution switch. In an overcurrent or short-circuit condition the junction temperature rises. When the die temperature rises to approximately 140°C due to overcurrent conditions, the internal thermal sense circuitry turns off the switch, thus preventing the device from damage. Hysteresis is built into the thermal sense, and after the device has cooled approximately 10 degrees, the switch turns back on. The switch continues to cycle off and on until the fault is removed. The open-drain false reporting output ($\overline{OC}$x) is asserted (active low) when an overtemperature shutdown or overcurrent occurs.

Undervoltage Lockout

A voltage sense circuit monitors the input voltage. When the input voltage is below approximately 2 V, a control signal turns off the power switch.

PACKAGING INFORMATION

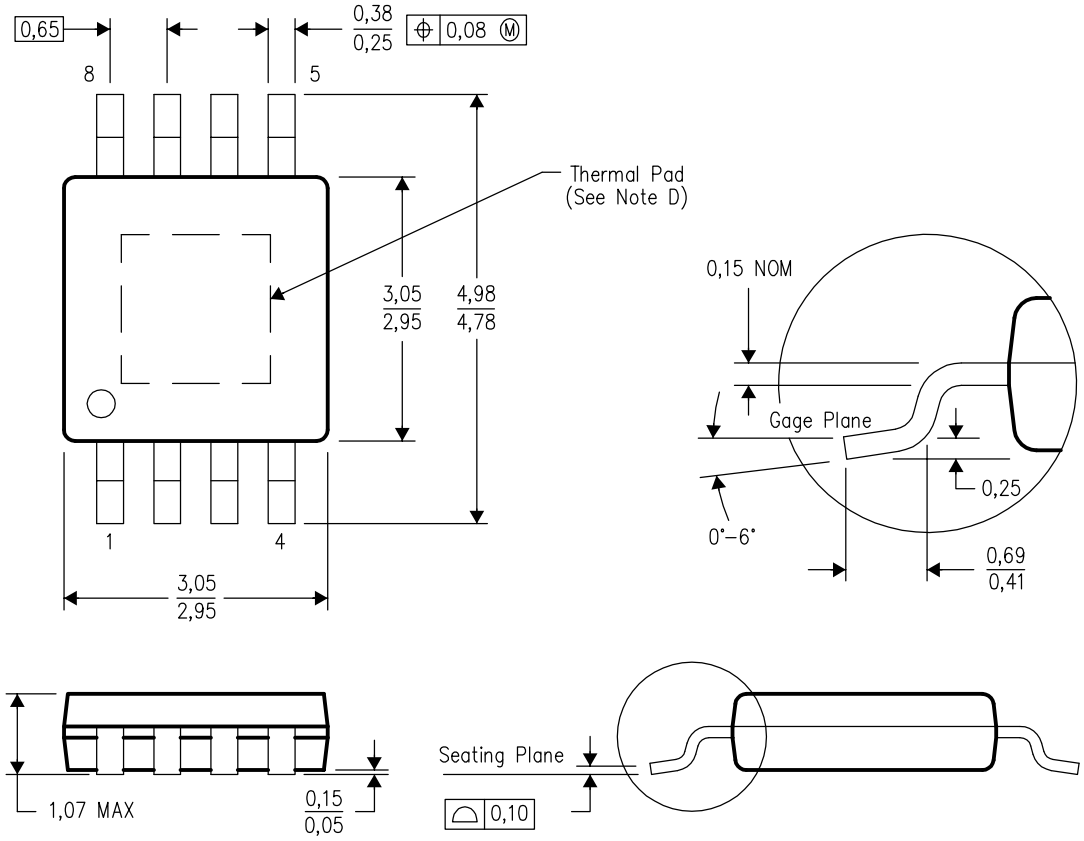
Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
TPS2060DGN	ACTIVE	MSOP-Power PAD	DGN	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2060DGNG4	ACTIVE	MSOP-Power PAD	DGN	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2060DGNR	ACTIVE	MSOP-Power PAD	DGN	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2060DGNRG4	ACTIVE	MSOP-Power PAD	DGN	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2064DGN	ACTIVE	MSOP-Power PAD	DGN	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2064DGNG4	ACTIVE	MSOP-Power PAD	DGN	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2064DGNR	ACTIVE	MSOP-Power PAD	DGN	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2064DGNRG4	ACTIVE	MSOP-Power PAD	DGN	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

DGN (S–PDSO–G8) PowerPAD™ PLASTIC SMALL–OUTLINE PACKAGE



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NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <http://www.ti.com>.
E. Falls within JEDEC MO–187

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