

FEATURES

- Low Power: $I_{CC} = 300\mu\text{A}$ Typical
- Designed for RS485 or RS422 Applications
- Single +5V Supply
- -7V to +12V Bus Common Mode Range
Permits $\pm 7\text{V}$ Ground Difference Between Devices on the Bus
- Thermal Shutdown Protection
- Power-Up/Down Glitch-Free Driver Outputs Permit Live Insertion or Removal of Package
- Driver Maintains High Impedance in Three-State or with the Power Off
- Combined Impedance of a Driver Output and Receiver Allows up to 32 Transceivers on the Bus
- 70mV Typical Input Hysteresis
- 28ns Typical Driver Propagation Delays with 5ns Skew
- Pin Compatible with the SN75180

APPLICATIONS

- Low Power RS485/RS422 Transceiver
- Level Translator

DESCRIPTION

The LTC491 is a low power differential bus/line transceiver designed for multipoint data transmission standard RS485 applications with extended common mode range (+12V to -7V). It also meets the requirements of RS422.

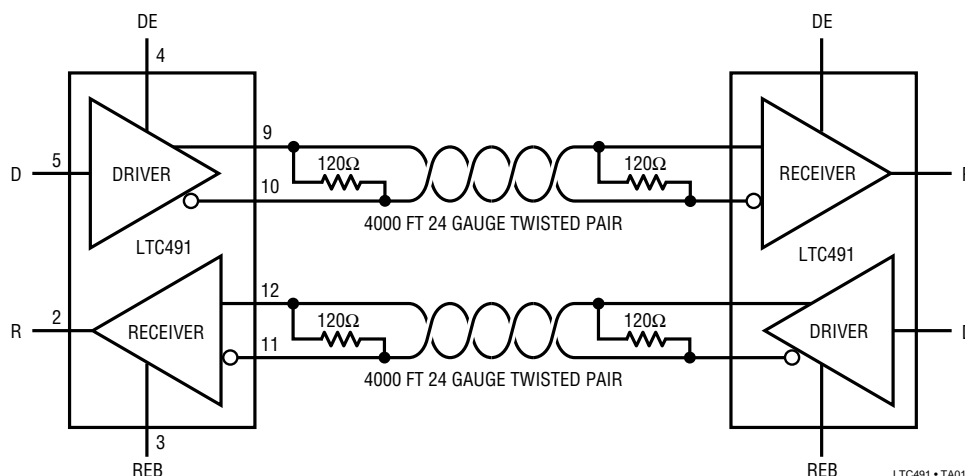
The CMOS design offers significant power savings over its bipolar counterpart without sacrificing ruggedness against overload or ESD damage.

The driver and receiver feature three-state outputs, with the driver outputs maintaining high impedance over the entire common mode range. Excessive power dissipation caused by bus contention or faults is prevented by a thermal shutdown circuit which forces the driver outputs into a high impedance state.

The receiver has a fail safe feature which guarantees a high output state when the inputs are left open.

Both AC and DC specifications are guaranteed from 0°C to 70°C and 4.75V to 5.25V supply voltage range.

TYPICAL APPLICATION

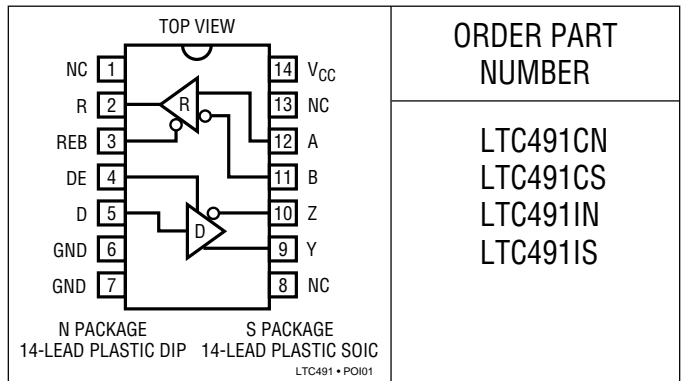


ABSOLUTE MAXIMUM RATINGS

(Note 1)

Supply Voltage (V_{CC})	12V
Control Input Voltages	-0.5V to $V_{CC} + 0.5V$
Control Input Currents	-50mA to 50mA
Driver Input Voltages	-0.5V to $V_{CC} + 0.5V$
Driver Input Currents	-25mA to 25mA
Driver Output Voltages	$\pm 14V$
Receiver Input Voltages	$\pm 14V$
Receiver Output Voltages	-0.5V to $V_{CC} + 0.5V$
Operating Temperature Range	
LTC491C	0°C to 70°C
LTC491I	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Lead Temperature (Soldering, 10 sec.)	300°C

PACKAGE/ORDER INFORMATION



ORDER PART NUMBER

LTC491CN
LTC491CS
LTC491IN
LTC491IS

Consult factory for Military grade parts.

DC ELECTRICAL CHARACTERISTICS

$V_{CC} = 5V \pm 5\%$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OD1}	Differential Driver Output Voltage (Unloaded)	$I_O = 0$	●		5	V
V_{OD2}	Differential Driver Output Voltage (With load)	$R = 50\Omega$; (RS422)	●	2		V
		$R = 27\Omega$; (RS485) (Figure 1)	●	1.5	5	V
ΔV_{OD}	Change in Magnitude of Driver Differential Output Voltage for Complementary Output States	$R = 27\Omega$ or $R = 50\Omega$ (Figure 1)	●		0.2	V
V_{OC}	Driver Common Mode Output Voltage		●		3	V
$\Delta V_{OC} $	Change in Magnitude of Driver Common Mode Output Voltage for Complementary Output States		●		0.2	V
V_{IH}	Input High Voltage	D, DE, REB	●	2.0		V
V_{IL}	Input Low Voltage		●		0.8	V
I_{IN1}	Input Current		●		± 2	μA
I_{IN2}	Input Current (A, B)	$V_{CC} = 0V$ or $5.25V$	●		1.0	mA
		$V_{IN} = 12V$ $V_{IN} = -7V$	●		-0.8	mA
V_{TH}	Differential Input Threshold Voltage for Receiver	$-7V \leq V_{CM} \leq 12V$	●	-0.2	0.2	V
ΔV_{TH}	Receiver Input Hysteresis	$V_{CM} = 0V$	●	70		mV
V_{OH}	Receiver Output High Voltage	$I_O = -4mA$, $V_{ID} = 0.2V$	●	3.5		V
V_{OL}	Receiver Output Low Voltage	$I_O = 4mA$, $V_{ID} = -0.2V$	●		0.4	V
I_{OZR}	Three-State Output Current at Receiver	$V_{CC} = \text{Max } 0.4V \leq V_O \leq 2.4V$	●		± 1	μA
I_{CC}	Supply Current	No Load; D = GND, or V_{CC}	●	300	500	μA
		Outputs Enabled Outputs Disabled	●	300	500	μA
R_{IN}	Receiver Input Resistance	$-7V \leq V_{CM} \leq 12V$	●	12		k Ω
I_{OSD1}	Driver Short Circuit Current, $V_{OUT} = \text{High}$	$V_O = -7V$	●	100	250	mA
I_{OSD2}	Driver Short Circuit Current, $V_{OUT} = \text{Low}$	$V_O = 12V$	●	100	250	mA
I_{OSR}	Receiver Short Circuit Current	$0V \leq V_O \leq V_{CC}$	●	7	85	mA
I_{OZ}	Driver Three-State Output Current	$V_O = -7V$ to $12V$	●	± 2	± 200	μA

SWITCHING CHARACTERISTICS

$V_{CC} = 5V \pm 5\%$

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
t_{PLH}	Driver Input to Output	$R_{DIFF} = 54\Omega$, $C_{L1} = C_{L2} = 100pF$ (Figures 2, 5)	●	10	30	50	ns
t_{PHL}	Driver Input to Output		●	10	30	50	ns
t_{SKEW}	Driver Output to Output		●		5		ns
t_r , t_f	Driver Rise or Fall Time		●	5	15	25	ns
t_{ZH}	Driver Enable to Output High	$C_L = 100pF$ (Figures 4, 6) S2 Closed	●		40	70	ns
t_{ZL}	Driver Enable to Output Low	$C_L = 100pF$ (Figures 4, 6) S1 Closed	●		40	70	ns
t_{LZ}	Driver Disable Time From Low	$C_L = 15pF$ (Figures 4, 6) S1 Closed	●		40	70	ns
t_{HZ}	Driver Disable Time From High	$C_L = 15pF$ (Figures 4, 6) S2 Closed	●		40	70	ns
t_{PLH}	Receiver Input to Output	$R_{DIFF} = 54\Omega$, $C_{L1} = C_{L2} = 100pF$ (Figures 2, 7)	●	40	70	150	ns
t_{PHL}	Receiver Input to Output		●	40	70	150	ns
t_{SKD}	$ t_{PLH} - t_{PHL} $ Differential Receiver Skew		●		13		ns
t_{ZL}	Receiver Enable to Output Low	$C_L = 15pF$ (Figures 3, 8) S1 Closed	●		20	50	ns
t_{ZH}	Receiver Enable to Output High	$C_L = 15pF$ (Figures 3, 8) S2 Closed	●		20	50	ns
t_{LZ}	Receiver Disable From Low	$C_L = 15pF$ (Figures 3, 8) S1 Closed	●		20	50	ns
t_{HZ}	Receiver Disable From High	$C_L = 15pF$ (Figures 3, 8) S2 Closed	●		20	50	ns

The ● denotes specifications which apply over the full operating temperature range.

Note 1: “Absolute Maximum Ratings” are those beyond which the safety of the device cannot be guaranteed.

Note 2: All currents into device pins are positive; all currents out of device pins are negative. All voltages are referenced to device ground unless otherwise specified.

Note 3: All typicals are given for $V_{CC} = 5V$ and Temperature = 25°C.

PIN FUNCTIONS

NC (Pin 1): Not Connected.

R (Pin 2): Receiver output. If the receiver output is enabled (REB low), then if $A > B$ by 200mV, R will be high. If $A < B$ by 200mV, then R will be low.

REB (Pin 3): Receiver output enable. A low enables the receiver output, R. A high input forces the receiver output into a high impedance state.

DE (Pin 4): Driver output enable. A high on DE enables the driver outputs, A and B. A low input forces the driver outputs into a high impedance state.

D (Pin 5): Driver input. If the driver outputs are enabled (DE high), then A low on D forces the driver outputs A low and B high. A high on D will force A high and B low.

GND (Pin 6): Ground Connection.

GND (Pin 7): Ground Connection.

NC (Pin 8): Not Connected.

Y (Pin 9): Driver output.

Z (Pin 10): Driver output.

B (Pin 11): Receiver input.

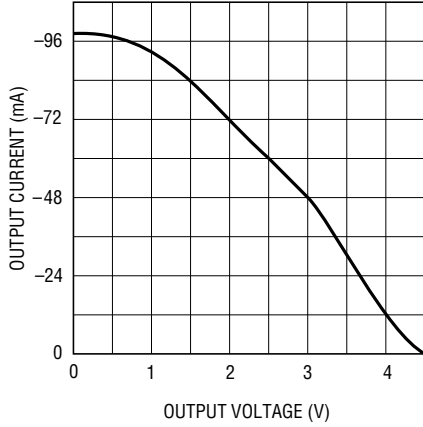
A (Pin 12): Receiver input.

NC (Pin 13): Not Connected.

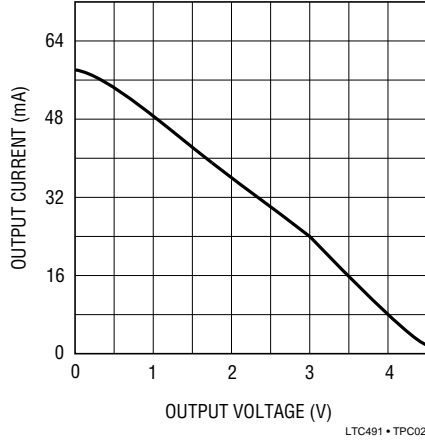
V_{CC} (Pin 14): Positive supply; $4.75V \leq V_{CC} \leq 5.25V$.

TYPICAL PERFORMANCE CHARACTERISTICS

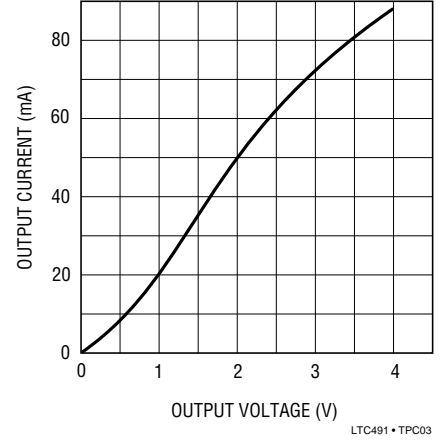
Driver Output High Voltage vs Output Current $T_A = 25^\circ\text{C}$



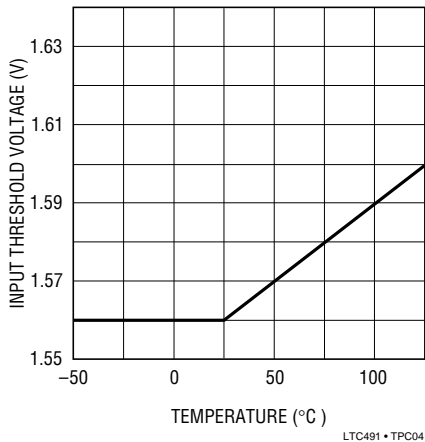
Driver Differential Output Voltage vs Output Current $T_A = 25^\circ\text{C}$



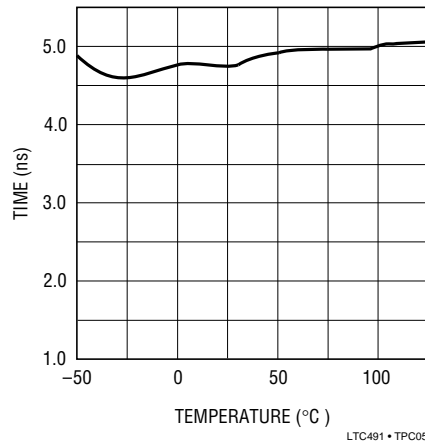
Driver Output Low Voltage vs Output Current $T_A = 25^\circ\text{C}$



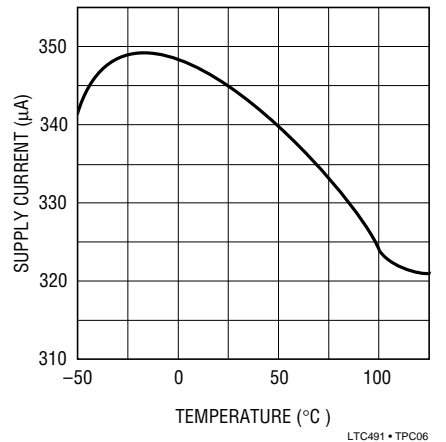
TTL Input Threshold vs Temperature



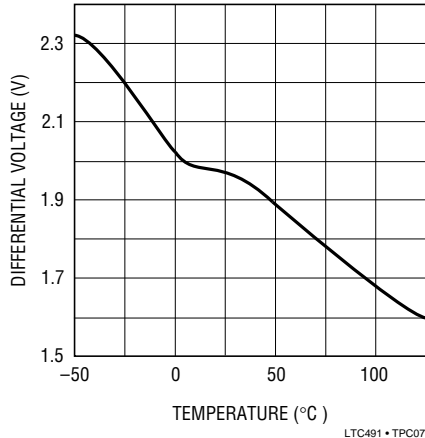
Driver Skew vs Temperature



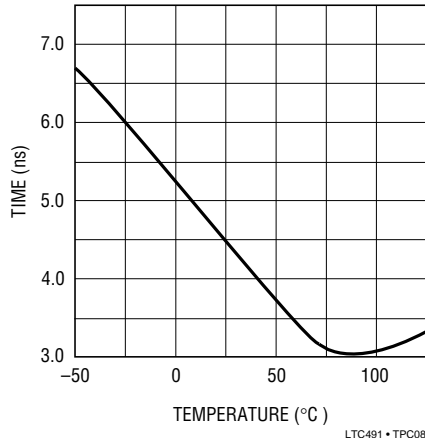
Supply Current vs Temperature



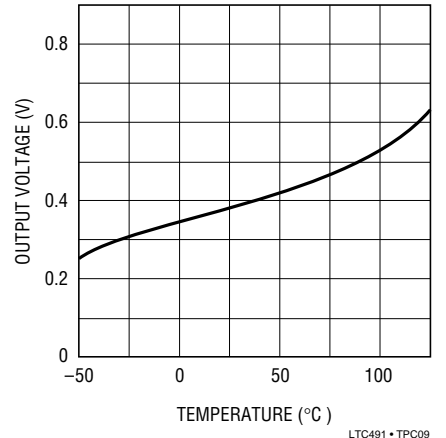
Driver Differential Output Voltage vs Temperature $R_0 = 54\Omega$



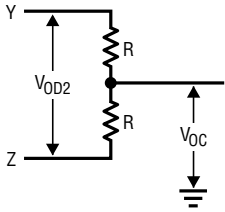
Receiver $|t_{PLH} t_{PHL}|$ vs Temperature



Receiver Output Low Voltage vs Temperature at $I = 8\text{mA}$

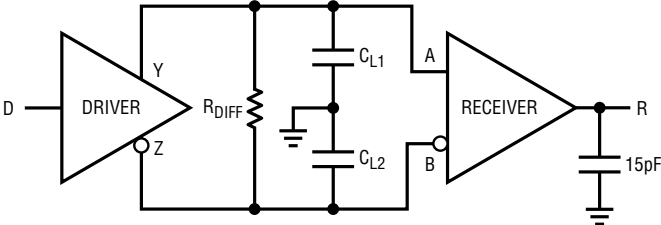


TEST CIRCUITS



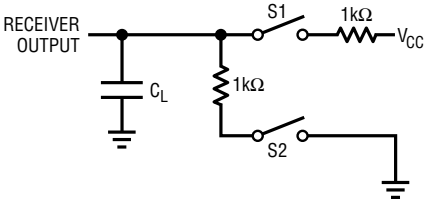
LTC491 • TA02

Figure 1. Driver DC Test Load



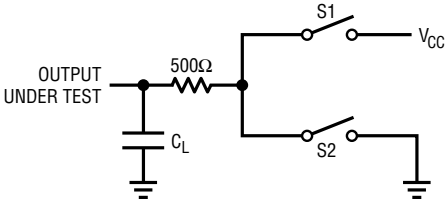
LTC491 • TA03

Figure 2. Driver/Receiver Timing Test Circuit



LTC491 • TA04

Figure 3. Receiver Timing Test Load



LTC491 • TA05

Figure 4. Driver Timing Test Load

SWITCHING TIME WAVEFORMS

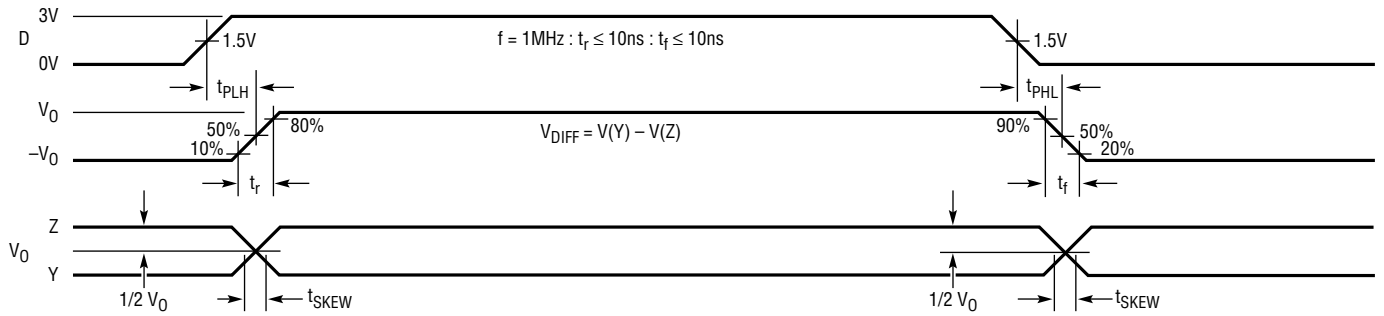


Figure 5. Driver Propagation Delays

LTC491 • TA06

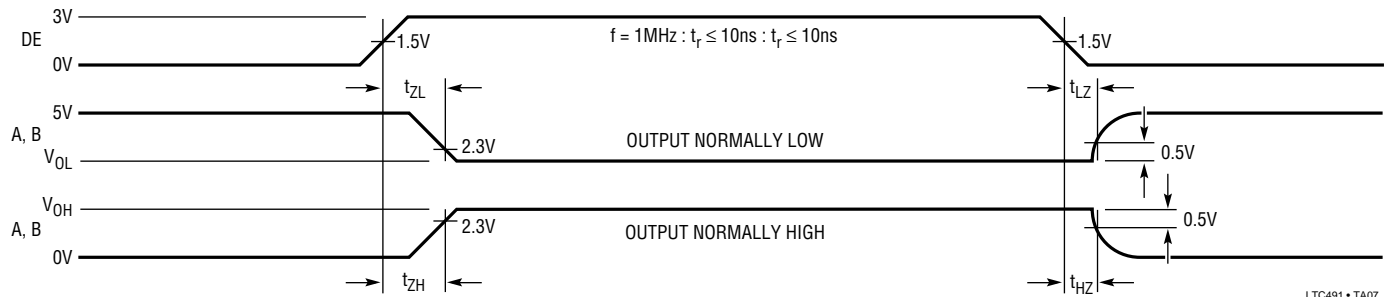


Figure 6. Driver Enable and Disable Times

LTC491 • TA07

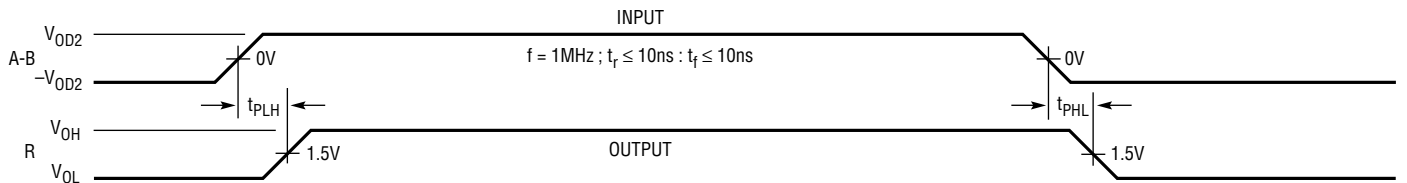


Figure 7. Receiver Propagation Delays

LTC491 • TA08

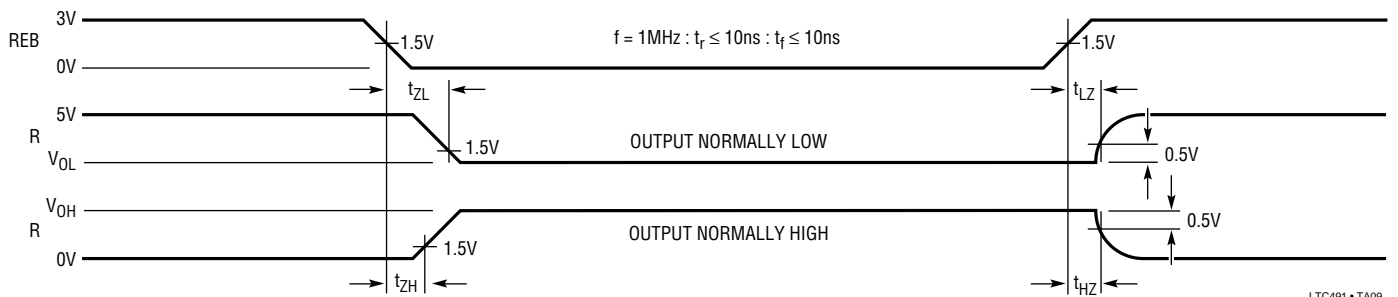


Figure 8. Receiver Enable and Disable Times

LTC491 • TA09

APPLICATIONS INFORMATION

Typical Application

A typical connection of the LTC491 is shown in Figure 9. Two twisted pair wires connect up to 32 driver/receiver pairs for full duplex data transmission. There are no restrictions on where the chips are connected to the wires, and it isn't necessary to have the chips connected at the ends. However, the wires must be terminated only at the ends with a resistor equal to their characteristic impedance, typically 120Ω . The input impedance of a receiver is

typically $20k\Omega$ to GND, or 0.6 unit RS-485 load, so in practice 50 to 60 transceivers can be connected to the same wires. The optional shields around the twisted pair help reduce unwanted noise, and are connected to GND at one end.

The LTC491 can also be used as a line repeater as shown in Figure 10. If the cable length is longer than 4000 feet, the LTC491 is inserted in the middle of the cable with the receiver output connected back to the driver input.

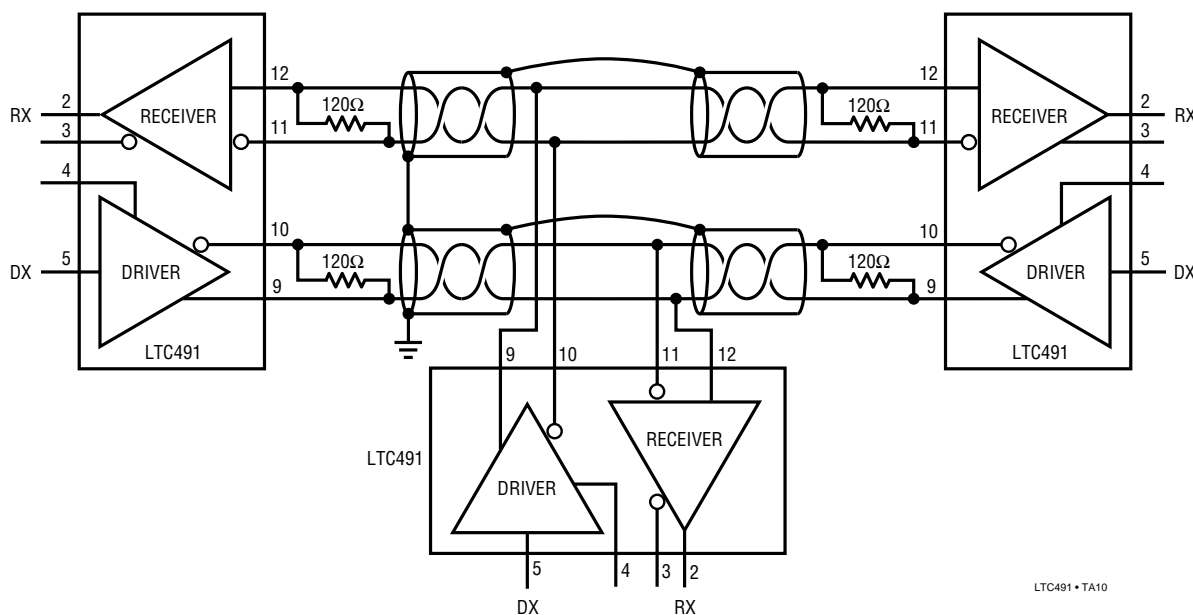


Figure 9. Typical Connection

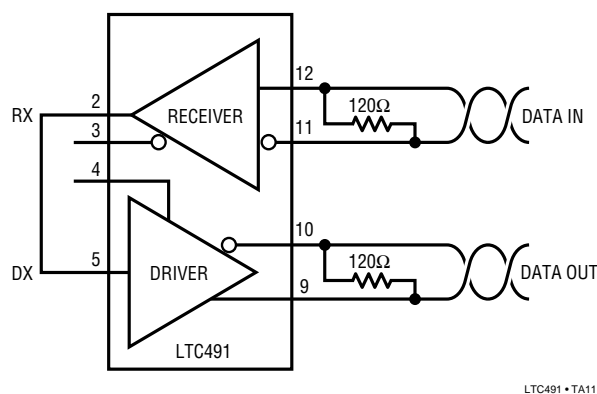


Figure 10. Line Repeater

APPLICATIONS INFORMATION

Thermal Shutdown

The LTC491 has a thermal shutdown feature which protects the part from excessive power dissipation. If the outputs of the driver are accidentally shorted to a power supply or low impedance source, up to 250mA can flow through the part. The thermal shutdown circuit disables the driver outputs when the internal temperature reaches 150°C and turns them back on when the temperature cools to 130°C. If the outputs of two or more LTC491 drivers are shorted directly, the driver outputs can not supply enough current to activate the thermal shutdown. Thus, the thermal shutdown circuit will not prevent contention faults when two drivers are active on the bus at the same time.

Cables and Data Rate

The transmission line of choice for RS485 applications is a twisted pair. There are coaxial cables (twinaxial) made for this purpose that contain straight pairs, but these are

less flexible, more bulky, and more costly than twisted pairs. Many cable manufacturers offer a broad range of 120Ω cables designed for RS485 applications.

Losses in a transmission line are a complex combination of DC conductor loss, AC losses (skin effect), leakage and AC losses in the dielectric. In good polyethylene cables such as the Belden 9841, the conductor losses and dielectric losses are of the same order of magnitude, leading to relatively low over all loss (Figure 11).

When using low loss cables, Figure 12 can be used as a guideline for choosing the maximum line length for a given data rate. With lower quality PVC cables, the dielectric loss factor can be 1000 times worse. PVC twisted pairs have terrible losses at high data rates (>100kBs), and greatly reduce the maximum cable length. At low data rates however, they are acceptable and much more economical.

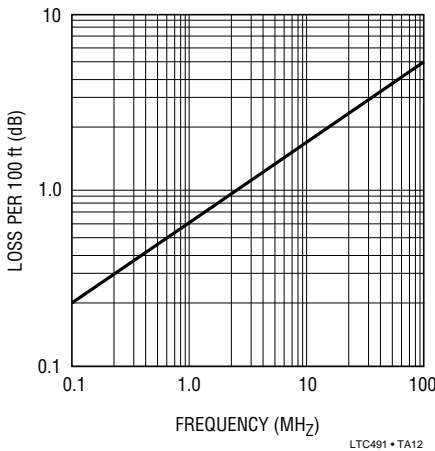


Figure 11. Attenuation vs Frequency for Belden 9481

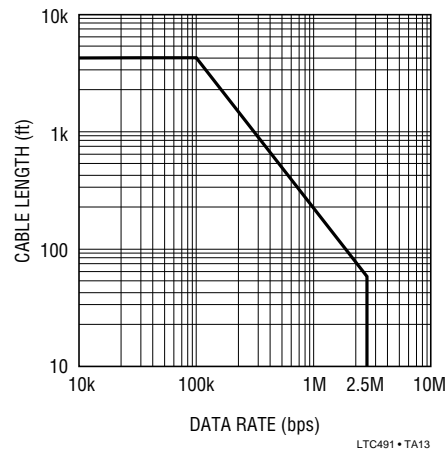


Figure 12. Cable Length vs Data Rate

APPLICATIONS INFORMATION

Cable Termination

The proper termination of the cable is very important. If the cable is not terminated with its characteristic impedance, distorted waveforms will result. In severe cases, distorted (false) data and nulls will occur. A quick look at the output of the driver will tell how well the cable is terminated. It is best to look at a driver connected to the end of the cable, since this eliminates the possibility of getting reflections from two directions. Simply look at the driver output while transmitting square wave data. If the cable is terminated properly, the waveform will look like a square wave (Figure 13).

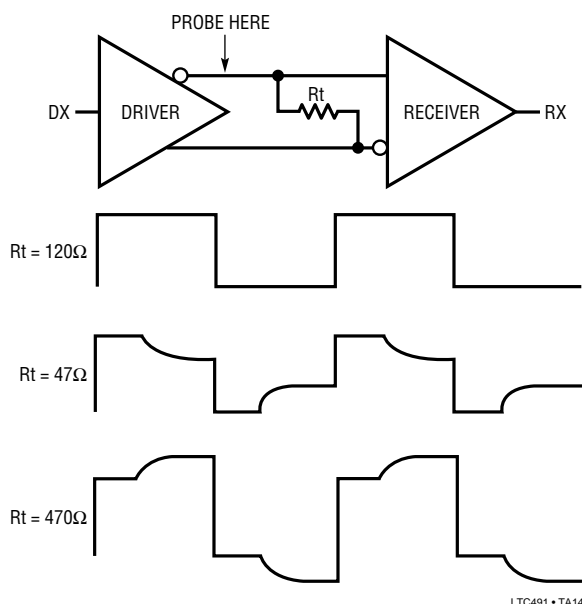


Figure 13. Termination Effects

If the cable is loaded excessively (47Ω), the signal initially sees the surge impedance of the cable and jumps to an initial amplitude. The signal travels down the cable and is reflected back out of phase because of the mistermination. When the reflected signal returns to the driver, the amplitude will be lowered. The width of the pedestal is equal to twice the electrical length of the cable (about 1.5ns/foot).

If the cable is lightly loaded (470Ω), the signal reflects in phase and increases the amplitude at the driver output. An input frequency of 30kHz is adequate for tests out to 4000 feet of cable.

AC Cable Termination

Cable termination resistors are necessary to prevent unwanted reflections, but they consume power. The typical differential output voltage of the driver is 2V when the cable is terminated with two 120Ω resistors, causing 33mA of DC current to flow in the cable when no data is being sent. This DC current is about 60 times greater than the supply current of the LTC491. One way to eliminate the unwanted current is by AC coupling the termination resistors as shown in Figure 14.

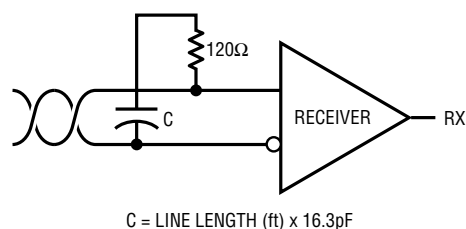


Figure 14. AC Coupled Termination

The coupling capacitor must allow high-frequency energy to flow to the termination, but block DC and low frequencies. The dividing line between high and low frequency depends on the length of the cable. The coupling capacitor must pass frequencies above the point where the line represents an electrical one-tenth wavelength. The value of the coupling capacitor should therefore be set at 16.3pF per foot of cable length for 120Ω cables. With the coupling capacitors in place, power is consumed only on the signal edges, and not when the driver output is idling at a 1 or 0 state. A 100nF capacitor is adequate for lines up to 4000 feet in length. Be aware that the power savings start to decrease once the data rate surpasses $1/(120\Omega \times C)$.

APPLICATIONS INFORMATION

Receiver Open-Circuit Fail-Safe

Some data encoding schemes require that the output of the receiver maintains a known state (usually a logic 1) when the data is finished transmitting and all drivers on the line are forced into three-state. The receiver of the LTC491 has a fail-safe feature which guarantees the output to be in a logic 1 state when the receiver inputs are left floating (open-circuit). However, when the cable is terminated with 120Ω , the differential inputs to the receiver are shorted together, not left floating. Because the receiver has about 70mV of hysteresis, the receiver output will maintain the last data bit received.

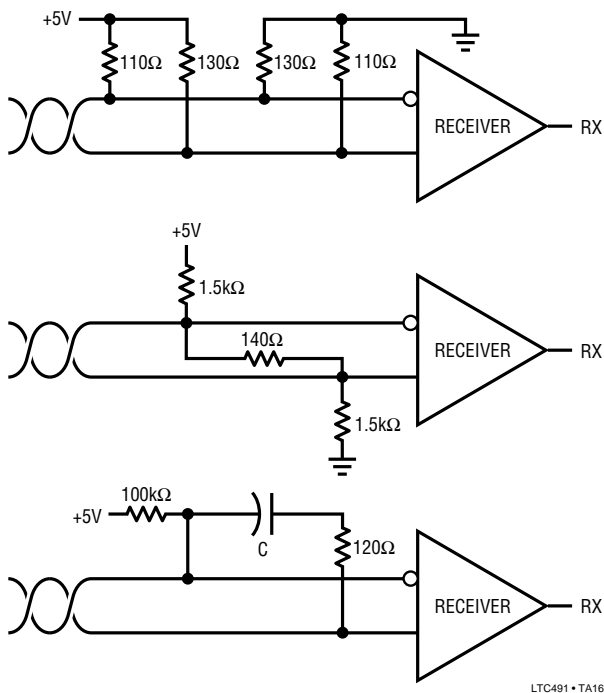


Figure 15. Forcing “0” When All Drivers are Off

The termination resistors are used to generate a DC bias which forces the receiver output to a known state, in this case a logic 0. The first method consumes about 208mW and the second about 8mW. The lowest power solution is to use an AC termination with a pull-up resistor. Simply swap the receiver inputs for data protocols ending in logic 1.

Fault Protection

All of LTC’s RS485 products are protected against ESD transients up to 2kV using the human body model (100pF, 1.5k Ω). However, some applications need more protection. The best protection method is to connect a bidirectional TransZorb from each line side pin to ground (Figure 16).

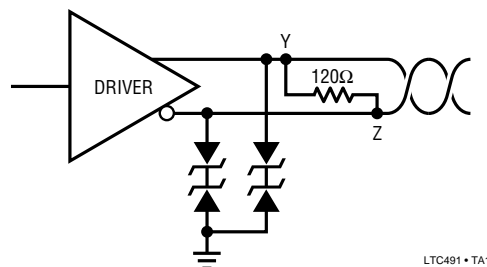
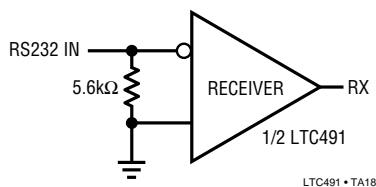


Figure 16. ESD Protection with TransZorbs

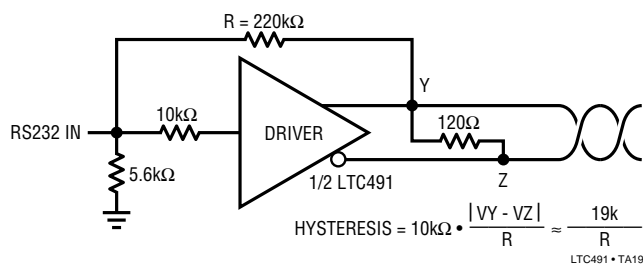
A TransZorb is a silicon transient voltage suppressor that has exceptional surge handling capabilities, fast response time, and low series resistance. They are available from General Semiconductor Industries and come in a variety of breakdown voltages and prices. Be sure to pick a breakdown voltage higher than the common mode voltage required for your application (typically 12V). Also, don’t forget to check how much the added parasitic capacitance will load down the bus.

TYPICAL APPLICATIONS

RS232 Receiver

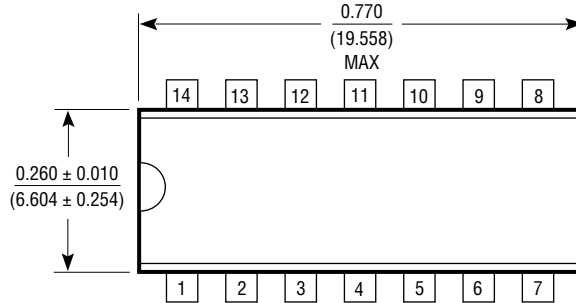


RS232 to RS485 Level Transistor with Hysteresis

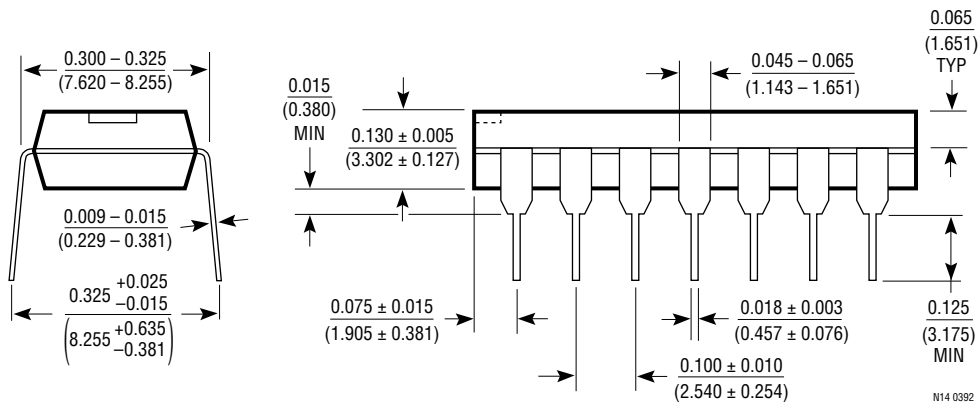


PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

**N Package
14-Lead Plastic DIP**

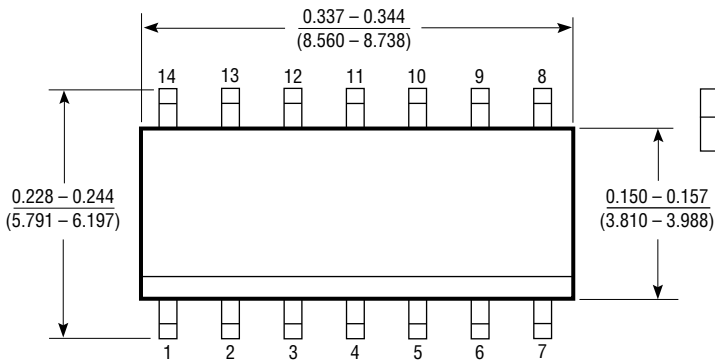


T_J MAX	θ_{JA}
100°C	90°C/W

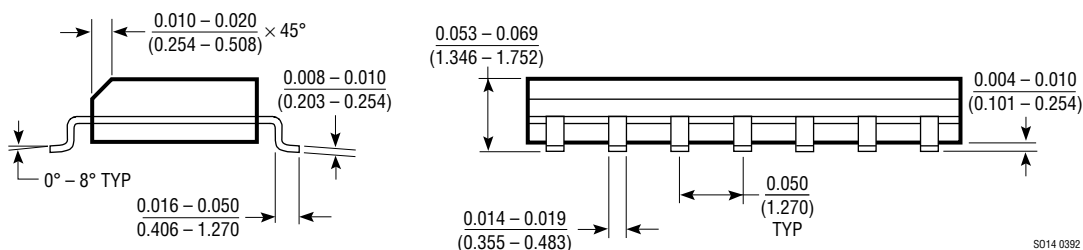


W14 0392

**S Package
14-Lead Plastic SOIC**



T_J MAX	θ_{JA}
100°C	110°C/W



S014 0392