

**40A, 600V, UFS Series N-Channel IGBTs**

The HGT1S20N60B3S, the HGTP20N60B3 and the HGTG20N60B3 are Generation III MOS gated high voltage switching devices combining the best features of MOSFETs and bipolar transistors. These devices have the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

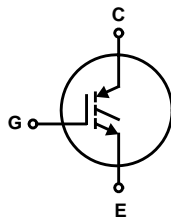
Formerly developmental type TA49050.

**Ordering Information**

PART NUMBER	PACKAGE	BRAND
HGTP20N60B3	TO-220AB	G20N60B3
HGT1S20N60B3S	TO-263AB	G20N60B3
HGTG20N60B3	TO-247	HG20N60B3

NOTE: When ordering, use the entire part number. Add the suffix 9A to obtain the TO-263AB in tape and reel, i.e., HGT1S20N60B3S9A.

**Symbol**

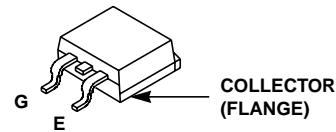


**Features**

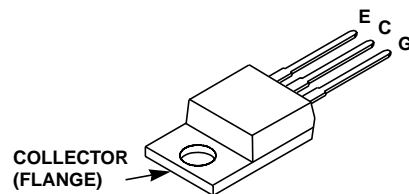
- 40A, 600V at  $T_C = 25^\circ\text{C}$
- 600V Switching SOA Capability
- Typical Fall Time . . . . . 140ns at  $150^\circ\text{C}$
- Short Circuit Rated
- Low Conduction Loss
- Related Literature
  - TB334 "Guidelines for Soldering Surface Mount Components to PC Boards"

**Packaging**

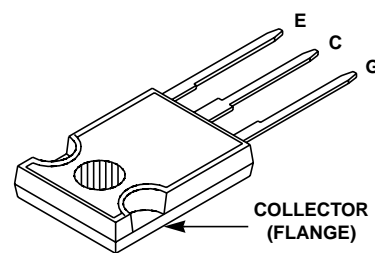
**JEDEC TO-263AB**



**JEDEC TO-220AB (ALTERNATE VERSION)**



**JEDEC STYLE TO-247**



**INTERSIL CORPORATION IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS**

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,587,713
4,598,461	4,605,948	4,620,211	4,631,564	4,639,754	4,639,762	4,641,162	4,644,637
4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690	4,794,432	4,801,986
4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606	4,860,080	4,883,767
4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951	4,969,027	

# HGT1S20N60B3S, HGTP20N60B3, HGTG20N60B3

## Absolute Maximum Ratings $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

	HGT1S20N60B3S HGTP20N60B3 HGTG20N60B3	UNITS
Collector to Emitter Voltage	600	V
Collector to Gate Voltage, $R_{GE} = 1\text{M}\Omega$	600	V
Collector Current Continuous		
At $T_C = 25^\circ\text{C}$	40	A
At $T_C = 110^\circ\text{C}$	20	A
Collector Current Pulsed (Note 1)	160	A
Gate to Emitter Voltage Continuous	$\pm 20$	V
Gate to Emitter Voltage Pulsed	$\pm 30$	V
Switching Safe Operating Area at $T_C = 150^\circ\text{C}$	30A at 600V	
Power Dissipation Total at $T_C = 25^\circ\text{C}$	165	W
Power Dissipation Derating $T_C > 25^\circ\text{C}$	1.32	$\text{W}/^\circ\text{C}$
Operating and Storage Junction Temperature Range	-40 to 150	$^\circ\text{C}$
Maximum Temperature for Soldering		
Leads at 0.063in (1.6mm) from Case for 10s	300	$^\circ\text{C}$
Package Body for 10s, see Tech Brief 334	260	$^\circ\text{C}$
Short Circuit Withstand Time (Note 2) at $V_{GE} = 15\text{V}$	4	$\mu\text{s}$
Short Circuit Withstand Time (Note 2) at $V_{GE} = 10\text{V}$	10	$\mu\text{s}$

*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

### NOTES:

1. Repetitive Rating: Pulse width limited by maximum junction temperature.
2.  $V_{CE} = 360\text{V}$ ,  $T_C = 125^\circ\text{C}$ ,  $R_G = 25\Omega$ .

## Electrical Specifications $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS	
Collector to Emitter Breakdown Voltage	$BV_{CES}$	$I_C = 250\mu\text{A}$ , $V_{GE} = 0\text{V}$	600	-	-	V	
Collector to Emitter Leakage Current	$I_{CES}$	$V_{CE} = BV_{CES}$	$T_C = 25^\circ\text{C}$	-	-	250	$\mu\text{A}$
			$T_C = 150^\circ\text{C}$	-	-	1.0	mA
Collector to Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = I_{C110}$ , $V_{GE} = 15\text{V}$	$T_C = 25^\circ\text{C}$	-	1.8	2.0	V
			$T_C = 150^\circ\text{C}$	-	2.1	2.5	V
Gate to Emitter Threshold Voltage	$V_{GE(TH)}$	$I_C = 250\mu\text{A}$ , $V_{CE} = V_{GE}$	3.0	5.0	6.0	V	
Gate to Emitter Leakage Current	$I_{GES}$	$V_{GE} = \pm 20\text{V}$	-	-	$\pm 100$	nA	
Switching SOA	SSOA	$T_C = 150^\circ\text{C}$ , $V_{GE} = 15\text{V}$ , $R_G = 10\Omega$ , $L = 45\mu\text{H}$	$V_{CE} = 480\text{V}$	100	-	-	A
			$V_{CE} = 600\text{V}$	30	-	-	A
Gate to Emitter Plateau Voltage	$V_{GEP}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$	-	8.0	-	V	
On-State Gate Charge	$Q_{G(ON)}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$	$V_{GE} = 15\text{V}$	-	80	105	nC
			$V_{GE} = 20\text{V}$	-	105	135	nC
Current Turn-On Delay Time	$t_{d(ON)I}$	$T_C = 150^\circ\text{C}$ $I_{CE} = I_{C110}$ $V_{CE} = 0.8 BV_{CES}$ $V_{GE} = 15\text{V}$ $R_G = 10\Omega$ $L = 100\mu\text{H}$	-	25	-	ns	
Current Rise Time	$t_{rl}$		-	20	-	ns	
Current Turn-Off Delay Time	$t_{d(OFF)I}$		-	220	275	ns	
Current Fall Time	$t_{fl}$		-	140	175	ns	
Turn-On Energy	$E_{ON}$		-	475	-	$\mu\text{J}$	
Turn-Off Energy (Note 3)	$E_{OFF}$		-	1050	-	$\mu\text{J}$	
Thermal Resistance	$R_{\theta JC}$		-	-	0.76	$^\circ\text{C}/\text{W}$	

### NOTE:

3. Turn-Off Energy Loss ( $E_{OFF}$ ) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ( $I_{CE} = 0\text{A}$ ). The HGT1S20N60B3S, HGTP20N60B3 and HGTG20N60B3 were tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss. Turn-On losses include diode losses.

Typical Performance Curves

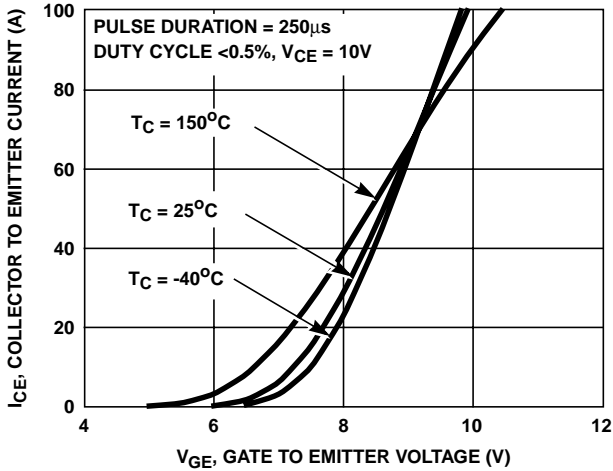


FIGURE 1. TRANSFER CHARACTERISTICS

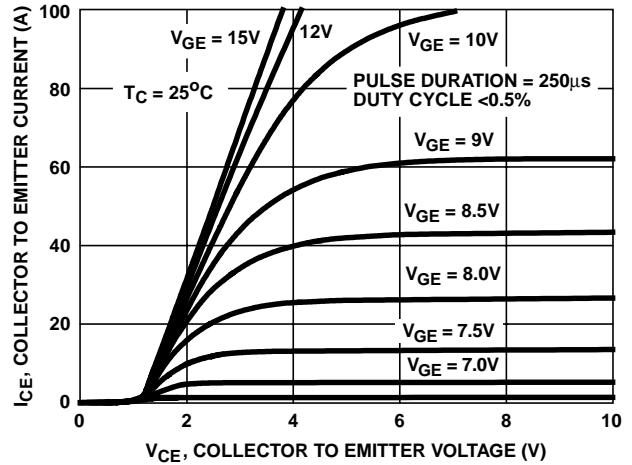


FIGURE 2. SATURATION CHARACTERISTICS

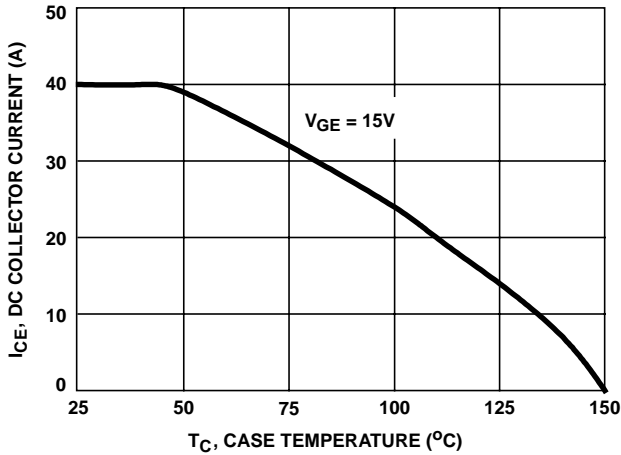


FIGURE 3. DC COLLECTOR CURRENT vs CASE TEMPERATURE

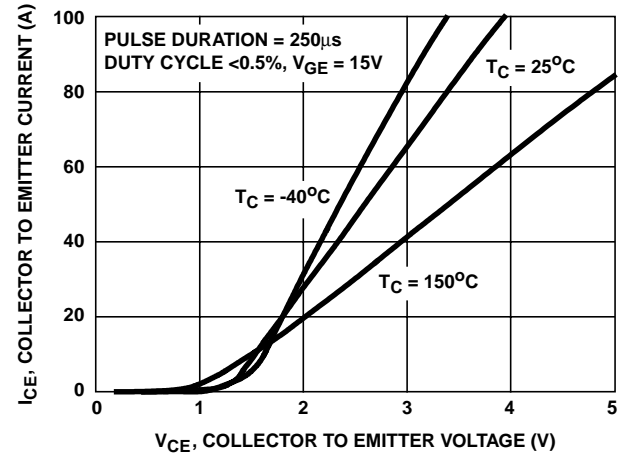


FIGURE 4. COLLECTOR TO EMITTER ON-STATE VOLTAGE

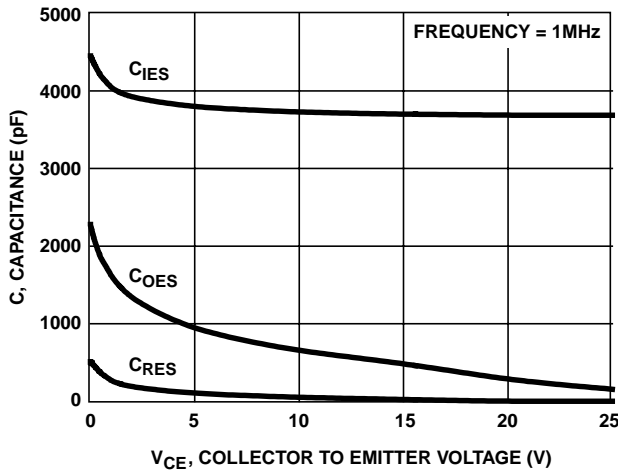


FIGURE 5. CAPACITANCE vs COLLECTOR TO EMITTER VOLTAGE

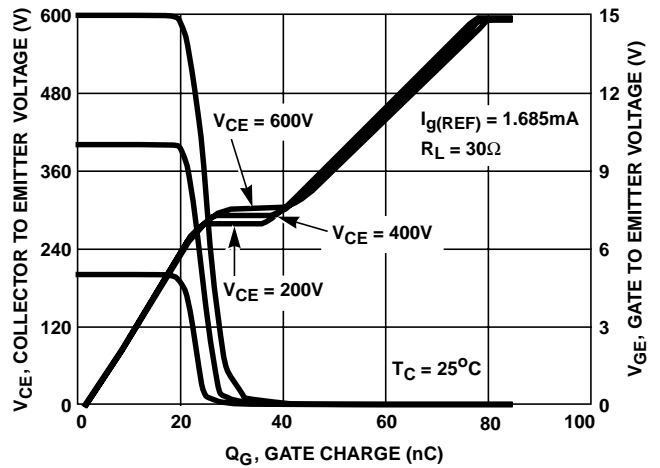


FIGURE 6. GATE CHARGE WAVEFORMS

Typical Performance Curves (Continued)

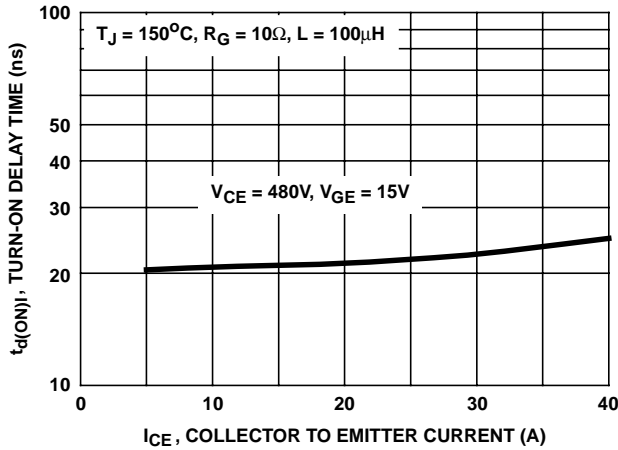


FIGURE 7. TURN-ON DELAY TIME vs COLLECTOR TO EMITTER CURRENT

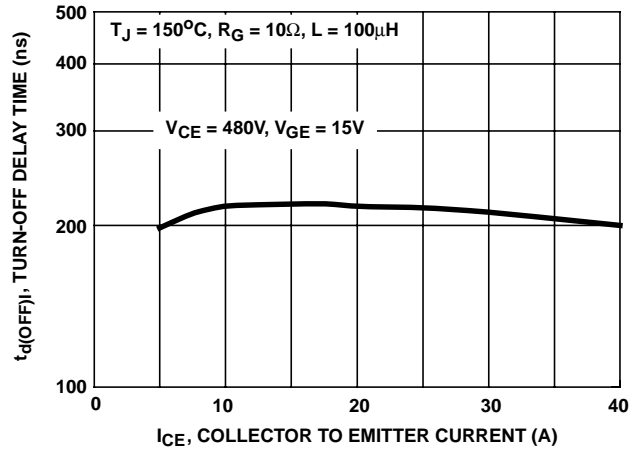


FIGURE 8. TURN-OFF DELAY TIME vs COLLECTOR TO EMITTER CURRENT

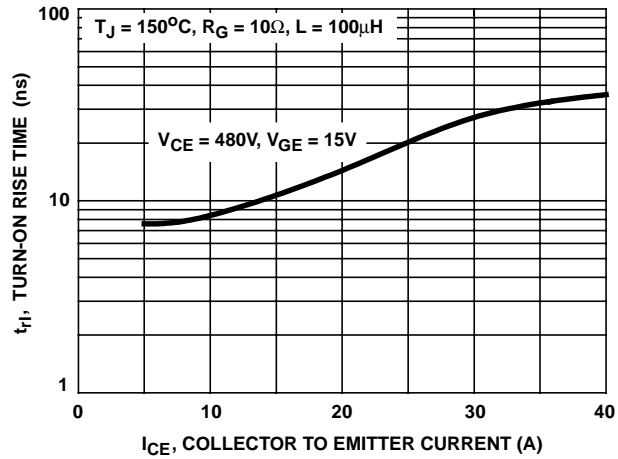


FIGURE 9. TURN-ON RISE TIME vs COLLECTOR TO EMITTER CURRENT

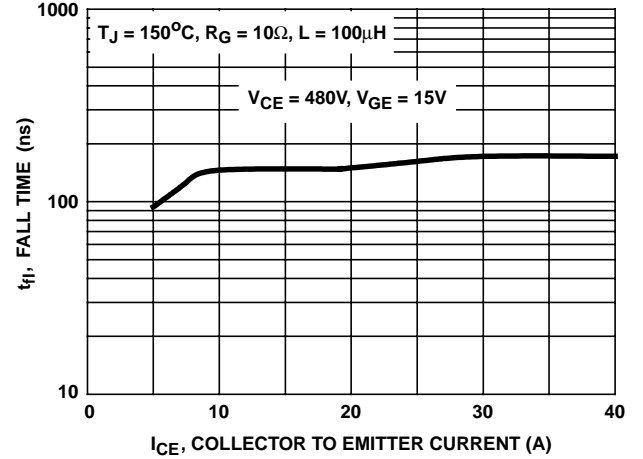


FIGURE 10. TURN-OFF FALL TIME vs COLLECTOR TO EMITTER CURRENT

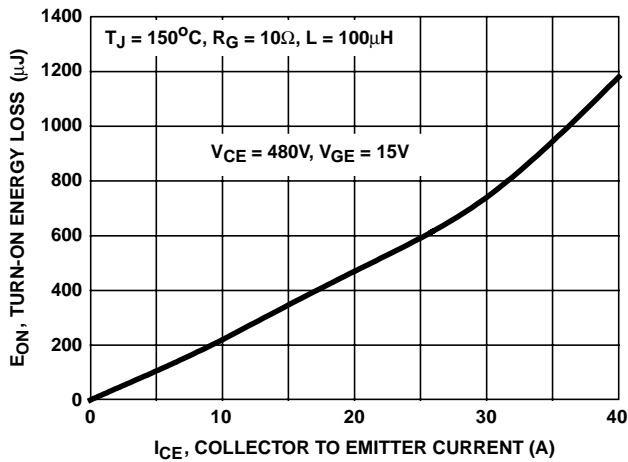


FIGURE 11. TURN-ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

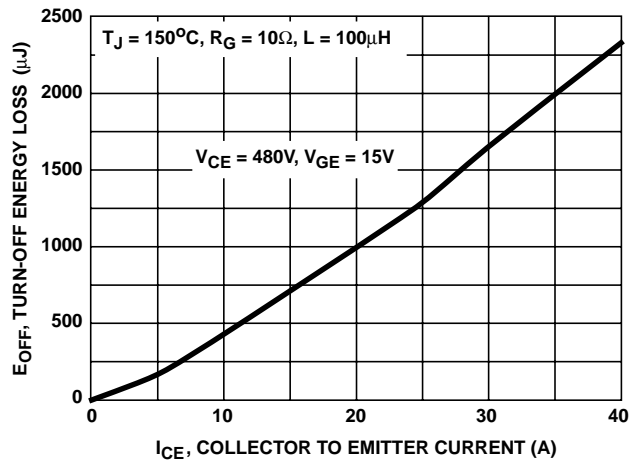


FIGURE 12. TURN-OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

Typical Performance Curves (Continued)

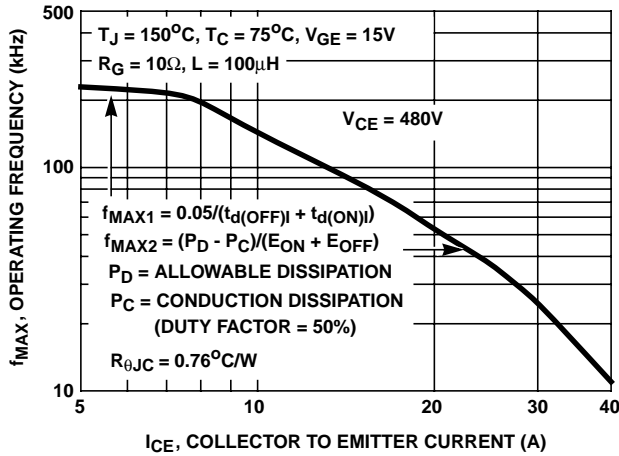


FIGURE 13. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

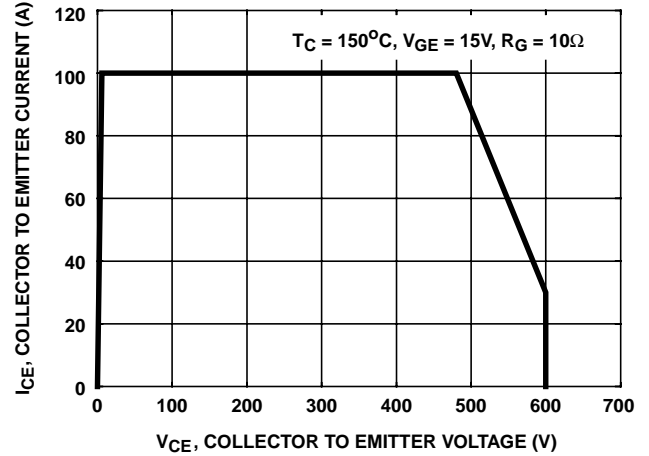


FIGURE 14. SWITCHING SAFE OPERATING AREA

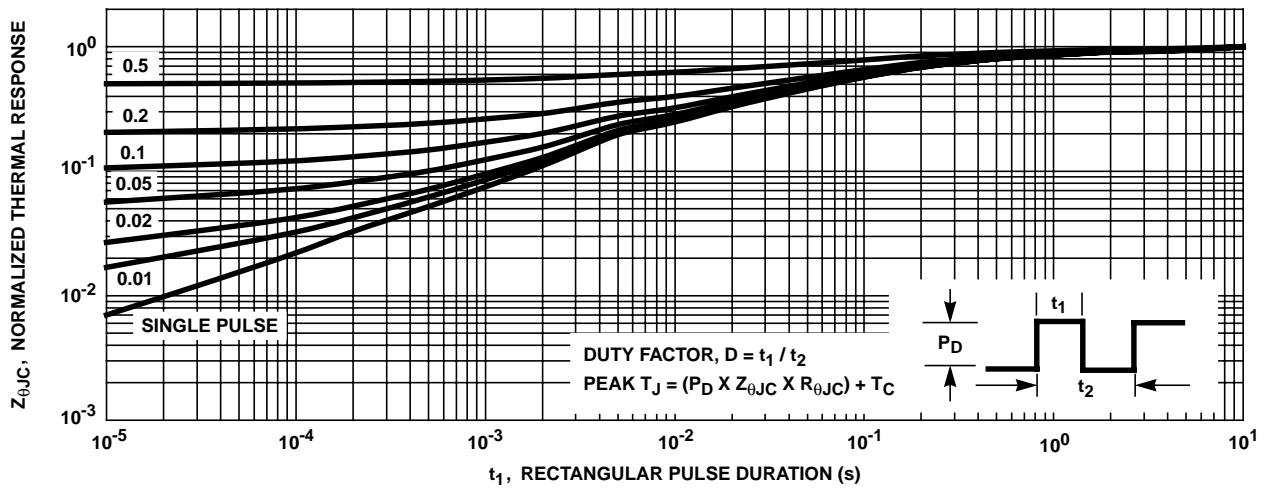


FIGURE 15. IGBT NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

Test Circuit and Waveform

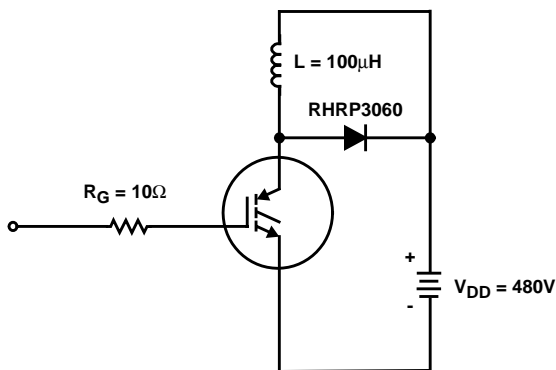


FIGURE 16. INDUCTIVE SWITCHING TEST CIRCUIT

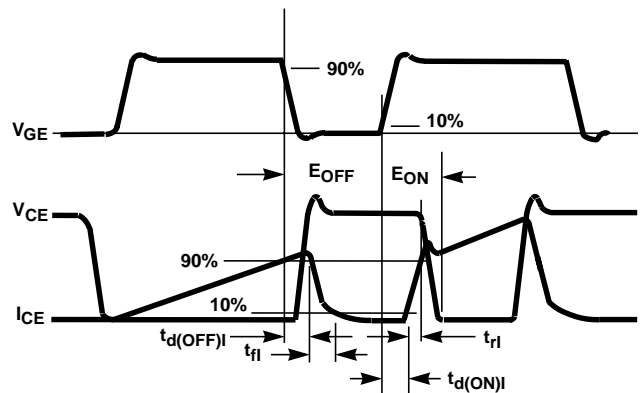


FIGURE 17. SWITCHING TEST WAVEFORMS

## Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic zener diode from gate to emitter. If gate protection is required an external zener is recommended.

## Operating Frequency Information

Operating frequency information for a typical device (Figure 13) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 4, 7, 8, 11 and 12. The operating frequency plot (Figure 13) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$  whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

$f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{d(OFF)I} + t_{d(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on- state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 17.

Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

$f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON})$ . The allowable dissipation ( $P_D$ ) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed  $P_D$ . A 50% duty factor was used (Figure 13) and the conduction losses ( $P_C$ ) are approximated by  $P_C = (V_{CE} \times I_{CE}) / 2$ .

$E_{ON}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 17.  $E_{ON}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

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